# **IDEA** League

Master of Science in Applied Geophysics Research Thesis

# Inversion of ERT data including uncertainties using the FE and RB methods

Niklas Krings

August 10, 2018

# Inversion of ERT data including uncertainties using the FE and RB methods

Master of Science Thesis

for the degree of Master of Science in Applied Geophysics at Delft University of Technology ETH Zürich RWTH Aachen University

by

Niklas Krings

August 10, 2018





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### IDEA LEAGUE JOINT MASTER'S IN APPLIED GEOPHYSICS

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### Abstract

<span id="page-8-0"></span>The inversion process of data sets acquired with the electrical resistivity tomography can be a computational expensive calculation, because it requires a lot of forward simulations. Since computational resources are limited, the inversion for large field surveys can require immense computing times. One approach for reducing the computing times of forward simulations is the reduced basis method. It splits the problem into a parameter-dependent and parameterindependent part, which allows a precomputation of the parameter-independent part to speed up the calculation. Since the method of reduced basis has not been applied for the problem of electrical resistivity tomography so far, the question arises if and how it is applicable for this problem. Therefore, I performed forward simulations of the electrical potential distribution using the finite elements method and simulations using the reduced basis method for comparing the results. The methods give similar results, but unfortunately deviations above  $1 \times 10^{-4}$  occurred, while the error tolerance of the reduced basis method is set to  $1 \times 10^{-5}$ . Since the reduced basis method has been successfully applied on other applications, I assume that the problem is in my reduced basis simulation and can be solved by more fine tuning. By comparing the computing times of the simulations, a speed-up of 109.512 is achieved by using the reduced basis method. Furthermore, the pay-off is reached after 8.803 simulations. During the project, I noticed that only considered the resistivities of the subsurface as a varying parameter in the reduced basis is not sufficient. The source position needs to be a variable parameter in simulations of the electrical resistivity tomography, as well. This is necessary to tap the full potential of the reduced basis method in the application of the electrical resistivity tomography. The reduced basis method gives significant speed-ups and therefore, it has the potential to reduce the computing times of electrical resistivity tomography inversions.

# Acknowledgements

<span id="page-10-0"></span>First of all, I want to thank all the people who have contributed to this project. Special thanks goes to Prof. Florian Wellmann for some inspiring discussions regarding my research. Furthermore, I want to thank my local supervisors Denise Degen and Rhea von Bülow for their support and feedback during the my research and the writing process. Additionally, I want to thank the institute for Computational Geoscience and Reservoir Engineering of the RWTH Aachen University for the very supportive and inspiring working environment. Last but not least, I want to thank my family and friends for letting me forget my thesis from time to time.

RWTH Aachen University Niklas Krings August 10, 2018

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# Acronyms

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# Chapter 1

### Introduction

<span id="page-18-0"></span>In geophysics, the limitations of the currently available computational resources have been an issue since the beginning of computer-aided calculations. And the demand for more computational resources probably will never be satisfied, because the processing of measured data and the modelling or simulation of physical systems can always be extended in the level of detail. Furthermore, there exist applications, that require the problem to be solved multiple times with varying parameters like an uncertainty quantification or a sensitivity analysis. Another intention to reduce the computational effort is make such calculations possible on mobile computers. This offers the possibility to perform data processing, which requires forward simulations, quickly in the field to evaluate the result. This could help to optimize the measurements and their parameters for obtaining the best possible result. Additionally, such faster computation enables real time processing of measured data. All in all, the need to reduce the computing time of calculations is omnipresent. One option for speeding up simulations is the reduced basis [\(RB\)](#page-16-1) method. The idea of the method is to separate the problem into a parameter-dependent and a parameter-independent part. This allows the parameter-independent part to be precomputed and the computation of the problem with a different parameter can be performed faster.

In this thesis, the [RB](#page-16-1) method will be applied to Electrical Resistivity Tomography [\(ERT\)](#page-16-4) data. [ERT](#page-16-4) surveys are used for obtaining a model of the near subsurface based on its resistivity. At first, synthetic data are generated and inverted. The purpose of this back-andforth-process is to investigate the induced uncertainties by the inversion process with the aid of an uncertainty quantification. On the base of the inversion result a model is generated, which is used for forward simulations using on the one hand the finite element [\(FE\)](#page-16-2) method and the [RB](#page-16-1) method on the other hand. In the original plan of this project, the final step was to perform with these two forward simulations an inversion and the uncertainty quantification. Unfortunately, the limited time did not allow to perform either an inversion nor an uncertainty quantification. With these large computations the required computational effort of the two methods should have been compared. This should have given proper answers on the question, how the [RB](#page-16-1) method can be useful for [ERT](#page-16-4) problems.

The inversion of [ERT](#page-16-4) data is a process, which requires multiple forward simulations. It is the most crucial step in the processing of measured [ERT](#page-16-4) data. The measurements contain only apparent resistivities and the inversion is necessary for obtaining the actual resistivities. In this process multiple forward simulations are performed. Commonly, the inversion is done with [FE](#page-16-2) inversion as, for example, implemented in the software package Boundless Electrical Resistivity Tomography [\(BERT\)](#page-16-3) [Rücker et al., 2006][Günther et al., 2006]. This state of the art method is described in [Rücker et al., 2006] and [Günther et al., 2006]. The forward simulations there are done using the [FE](#page-16-2) method. Other approaches use the finite difference [\(FD\)](#page-16-5) method (see [\[Wei and Versteeg, 2008\]](#page-63-0)). Wei also shows the need for an improved computational efforts for the inversion of [ERT](#page-16-4) data. There, they implement an approach to speed up the computation by splitting each simulated dipole-dipole measurement into two measurements using a single current electrode and combining the results [\[Wei and Versteeg, 2008\]](#page-63-0).

The [RB](#page-16-1) method has previously been applied to thermal conduction problems and elasticity problems (see [\[Hesthaven et al., 2016\]](#page-61-1), [\[Prud'Homme et al., 2002\]](#page-62-2) and [\[Boyaval et al., 2009\]](#page-60-1)). Additionally, Casenave (2015) applied the [RB](#page-16-1) method on aeroacoustic problems (see [\[Casenave et al., 2015\]](#page-60-2)). Other applications are transport problems in biomedical research. So, Rozza applied the [RB](#page-16-1) method for simulations for the flow control of bypass configurations (see [\[Rozza, 2005b\]](#page-62-3) and [\[Rozza, 2005a\]](#page-62-4)). Inverse problems also have been addressed using the [RB](#page-16-1) method. Lassila applied the [RB](#page-16-1) method in cardiovascular mathematics by replacing the finite element solution by a reduced-basis approximation [\[Lassila et al., 2013\]](#page-61-3). Manzoni utilized the [RB](#page-16-1) method to obtain an efficient numerical solution of statistical inverse problems in blood flows [\[Manzoni et al., 2014\]](#page-62-5). Furthermore, the [RB](#page-16-1) method has been applied to electromagnetics. On the one hand Maxwell's equations were modelled with the aid of the [RB](#page-16-1) method [\[Chen et al., 2009\]](#page-61-4)[\[Chen et al., 2010\]](#page-61-5). On the other hand it has been implemented for the electric field integral equation for parameterized scattering problems [\[Fares et al., 2011\]](#page-61-6)[\[Hesthaven et al., 2012\]](#page-61-7). The research done on the [RB](#page-16-1) method itself focuses mostly on the convergence (see [\[Binev et al., 2011\]](#page-60-3) and [\[Buffa et al., 2012\]](#page-60-4)) and improvement of the error bounds (see [\[Veroy et al., 2003\]](#page-62-6)).

Uncertainty quantification is a useful tools for analysing data. It has no limitation to a certain application and can be applied to data form various fields of science. For example, the method is used for investigating the simulations for predicting the outcome of medical surgeries (see [\[Schiavazzi et al., 2016\]](#page-62-7)). Another application for uncertainty quantification is for example image reconstruction. So, Bardsley tried to reconstructed blurred images using the Markov Chain Monte Carlo [\(MCMC\)](#page-16-6) method (see [\[Bardsley, 2012\]](#page-60-5)). Furthermore, it has been used for data of orbits of extrasolar planets. The uncertainty quantification is used to determine which simulation describes the measured radial velocities of extrasolar planetary systems (see [\[Ford, 2005\]](#page-61-8)). Therefore, Ford applies the [MCMC](#page-16-6) uncertainty quantification. [MCMC](#page-16-6) is a commonly used method and is a Bayesian approach [\[Adams et al., 2014\]](#page-60-0). But due to issues of efficiency, alternative method evolved like the Differential Evolution Adaptive Metropolis [\(DREAM\)](#page-16-7) method, which tries optimize the performance of the [MCMC](#page-16-6) method [\[Vrugt et al., 2009\]](#page-62-8). Since uncertainty quantifications require multiple forward simulations and efficiency is an issue, the [RB](#page-16-1) method can be helpful tool to speed up the computation of uncertainty quantifications.

At the beginning of this thesis, I will focus on the theory behind the applications and the methods themselves. At first geoelectric surveys and especially the [ERT](#page-16-4) will be explained in Section [2-1,](#page-22-1) followed by the inversion process. The next section describes the [FE](#page-16-2) approximation, with its basis functions in Section [2-2.](#page-28-0) The [RB](#page-16-1) method will be explained in Section [2-3.](#page-31-0) Additionally, I will explain the theory of uncertainty quantification in Section [2-4.](#page-36-0) After that, the general workflow of the thesis will be illustrated in Section [3-1](#page-38-1) and the used model is described in Section [3-2.](#page-40-0) Additionally, the employed software and their techniques will be discussed in Section [3-3.](#page-41-0) The results will be presented in Chapter [4.](#page-44-0) Furthermore, I will discuss my results in Chapter [5.](#page-56-0)

### Chapter 2

# Theory

<span id="page-22-0"></span>At the beginning, I describe the essentials of the theory in the following. The theory behind the electrical resistivity tomography [\(ERT\)](#page-16-4) is described in Section [2-1.](#page-22-1) Furthermore, the finite element [\(FE\)](#page-16-2) method is explained in Section [2-2,](#page-28-0) followed by the reduced basis [\(RB\)](#page-16-1) method. Finally, the theory behind uncertainty quantification is given in Section [2-4.](#page-36-0)

### <span id="page-22-1"></span>2-1 Electrical Resisitivity Tomography

In this chapter, the principles of [ERT](#page-16-4) will be explained. At first, the electrical properties of the Earth are in the focus (Section [2-1-1\)](#page-22-2), followed by the concept (Section [2-1-2\)](#page-23-0) and the common survey arrays of [ERT](#page-16-4) (Section [2-1-3\)](#page-25-0). In the last section the inversion of [ERT](#page-16-4) data is explained (Section [2-1-4\)](#page-26-0).

### <span id="page-22-2"></span>2-1-1 Electrical properties of the Earth

The Earth and its rocks have electrical properties as any other material. Therefore, it is also possible to investigate the subsurface by exploiting variations in its electrical properties. Most geoelectrical applications search either for differences in the electrical resistivity  $\rho$  or the electrical permittivity  $\epsilon$ . Surveys investigating the permittivity use usually electromagnetic methods, while direct current measurements are used to survey  $\rho$ . Anomalies in the subsurface with high contrasts in these properties to its surroundings can be well surveyed. Therefore, the most suitable investigation objects are orebodies and groundwater. The orebodies usually reveal a high contrast due to its high conductive metals. For example, sulfide ores of iron, copper and nickel. In the groundwater, solved ions show a significant conducting property. So, the groundwater appears as an electrolyte [\[Lowrie, 2007\]](#page-62-0). Most surveys use active sources, but there are also methods utilizing natural currents, also telluric currents. Their occurrence has two origins. Electrical currents in the ionosphere, originated from solar emissions, induce electromagnetically currents in the Earth's crust and upper mantle. Furthermore, electrochemical reactions in the subsurface can generate currents. Distinguishing

<span id="page-23-1"></span>

Figure 2-1: Conductivities and resistivities of common rocks, soils and ores. From [\[Lowrie, 2007\]](#page-62-0)

different rock types can be challenging, because the ranges of the surveyed property, e.g. resistivity, can overlap for different rock types. An overview of resistivities  $\rho$  and conductivities  $\sigma$  of common rocks, soils and ores is given in Figure [2-1.](#page-23-1) Furthermore, the resistivity of porous sediments and sedimentary rocks can vary significantly due to the porosity and the filling of pores. Because there are no unique resistivities for certain rocks, additional information from other survey methods are helpful for identifying different rock units [\[Lowrie, 2007\]](#page-62-0).

#### <span id="page-23-0"></span>2-1-2 Geoelectric surveys and electrical resistivity tomography

For a better understanding about geoelectrical surveys, the principles of the physics behind need to be clear. Most surveys investigate how resistive or conductive the subsurface is. So, the resistance R can be determined by the current I and the voltage U with the aid of Ohm's law [\[Lowrie, 2007\]](#page-62-0)

$$
U = RI.
$$
\n<sup>(2-1)</sup>

But the resistance is an extensive property and therefore not material characteristic. The intensive counterpart to the resistance is the resistivity  $\rho$ . Their relationship can be expressed

#### as [\[Lowrie, 2007\]](#page-62-0)

$$
R = \rho \frac{l}{A},\tag{2-2}
$$

where  $l$  denotes the length of the material sample and  $A$  represents its cross-sectional area. So, the resistivity is a material property and therefore more interesting for geophysical surveys, because it allows compared to the resistance the identification of various rock units. The conductivity  $\sigma$  is the reciprocal of the resistivity and is therefore expressed as  $\sigma = 1/\rho$ [\[Lowrie, 2007\]](#page-62-0).

Geoelectric surveys can either be active or passive. While in the case of active sources, an artificial current is injected into the ground, surveys using natural sources measure potentials of naturally occurring fields. These natural currents, also telluric currents, occur mostly due to electrochemical reactions within the subsurface. But there is no option for controlling the source, which can be strongly time dependent. Hence, interpretations are mostly only qualitative and not quantitative. However, the use of natural currents is inexpensive and fast. Active methods give usually better results, because the source is known and therefore also a quantitative interpretation is possible. Additionally the source is controllable and can be modified to optimize the survey for every specific case. Active sources of direct current measurements usually consist of two current electrodes, which are plugged into the ground. So, the Earth is part of the electric circuit. Commonly, a low-frequency alternating current is used for the source to have a direct current for a few seconds. The use of only direct current can cause accumulations of charges at the two potential electrodes, which leads to false signals. The potential electrodes measure the potential at different locations to give the gradient of the electric field [\[Lowrie, 2007\]](#page-62-0).

A simple option of geoelectric surveys is performing single measurements at certain locations of interest. But nowadays most surveys use the [ERT](#page-16-4) method. Instead of only four electrodes, an array of electrodes is employed for investigating an area of interest. Commonly evenspaced electrode arrays are utilized, but arbitrary configurations are also possible. This array can be a line for two-dimensional surveys or an areal array for three-dimensional surveys. An abundance of measurements are then performed with various combinations of two current and two potential electrodes. The selection of the four electrodes can be arbitrary, but it is common to use certain configurations of the electrodes. The most common configurations are explained in Section [2-1-3.](#page-25-0) For obtaining information from different depth levels, measurements with different electrode offset are performed. During the measurements, the flowing current is injected at the current electrodes, while the potential is measured at the potential electrodes. In combination with the known geometry, the apparent resistivities  $\rho_a$  can be calculated with the aid of Equation [2-3](#page-24-0) [\[Lowrie, 2007\]](#page-62-0)

<span id="page-24-0"></span>
$$
\rho_a = 2\pi \frac{\Delta \varphi}{I} \left[ \left( \frac{1}{r_{AC}} - \frac{1}{r_{CB}} \right) - \left( \frac{1}{r_{AD}} - \frac{1}{r_{DB}} \right) \right]^{-1}.
$$
 (2-3)

 $\Delta\varphi$  denotes the measured potential difference, I the current and r the distance between the two electrodes given in its index. The indices  $A, B, C$  and  $D$  refer to the four electrodes. A and  $B$  are the current electrodes, while  $C$  and  $D$  denote the potential electrodes. Note, the resistivity information from each measurement is gathered along the whole path of the current through the subsurface and not at a single point. Therefore, the calculated resistivity is only apparent. Hence, the measured apparent resistivity is the resistivity of a homogeneous half-space. Note that for the special case of a homogeneous half-space, the apparent resistivity

is the true resistivity. For any other case, the true resistivities can be obtained by performing an inversion, which is described in Section [2-1-4](#page-26-0) [\[Lowrie, 2007\]](#page-62-0).

#### <span id="page-25-0"></span>2-1-3 Common four-electrode configurations

The configuration of the electrodes, which describes the ordering of the four electrodes and their respective distances, can be chosen arbitrarily. But there are some specific configurations, which are commonly used. These are configurations are therefore well known and established. Using these has the advantage of comparability between different measurements. Furthermore, these configurations simplify the calculation of the apparent resistivity in Equa-tion [2-3,](#page-24-0) because the distances  $r$  between the electrodes follow defined schemes.

The most common electrode configurations are namely the Dipole-Dipole, the Wenner and the Schlumberger configuration. They are depicted in Figure [2-2.](#page-26-1) The Wenner configuration sets  $r_{AC} = r_{DB} = a$  and  $r_{AD} = r_{BC} = 2a$ , where a is an arbitrary distance. Therefore the equation for the apparent resistivity reduces to [\[Lowrie, 2007\]](#page-62-0)

$$
\rho_a = 2\pi a \frac{\Delta \varphi}{I}.\tag{2-4}
$$

While this equation becomes quite simple by the Wenner configuration, it is often inconvenient for [ERT](#page-16-4) field work. For example, the extend to larger offsets need to be in certain steps, because every distance needs to be the same a [\[Telford et al., 1990\]](#page-62-9). Furthermore, large offsets require long cables, especially between the two current electrodes, because in the Wenner configuration their distance is always larger than the distance between the potential electrodes. The Wenner configuration is compared to the other options more sensitive to lateral variations. This can be optimized when the current electrode offset is chosen so that the current flow is maximal in the depth where the most lateral variations are expected. This makes the Wenner configuraton suitable for steeply dipping interfaces such as dykes [\[Lowrie, 2007\]](#page-62-0).

The Schlumberger configuration is similar to the Wenner configuration, but it has less restriction to the electrode spacing. Both electrode pairs are required to have a common mid point, but their is no relation between the offset of the current electrodes and the potential electrodes. The distances are  $r_{AC} = r_{DB} = \frac{L-a}{2}$  $\frac{-a}{2}$  and  $r_{AD} = r_{CB} = \frac{L+a}{2}$  $\frac{+a}{2}$ , where L is the distance between the current electrodes and a is the distance between the potential electrodes. Usually, the current electrode distance is kept much larger than the potential electrode distance  $(L \gg a)$ . The apparent resistivity using the Schlumberger configuration can then be calculated by [\[Lowrie, 2007\]](#page-62-0)

$$
\rho_a = \frac{\pi \Delta \varphi}{4I} \left(\frac{L^2}{a}\right)^2.
$$
\n(2-5)

This configuration is often used for vertical electric sounding. It is a method for investigating resistivity variations in depth. Therefore, a common mid point is chosen and severals measurements with different offset of the electrodes are performed [\[Lowrie, 2007\]](#page-62-0).

In the Dipole-Dipole configuration, the current and the potential electrodes can be separated independently. The distance between the electrodes in both pairs need to be the same. But the distance between the two pairs can be arbitrarily. This enables an abundance of possibilities, which raised the popularity of the configuration for [ERT](#page-16-4) surveys. Here, the equation

#### (a) Wenner

<span id="page-26-1"></span>

(b) Schlumberger



(c) Dipole-Dipole



Figure 2-2: The different four-electrode configurations. a) Wenner, b) Schlumberger and c) Dipole-Dipole. Note that the potential difference is here denoted as  $\Delta V$ . After [\[Lowrie, 2007\]](#page-62-0)

for the apparent resistivity reduces to [\[Lowrie, 2007\]](#page-62-0)

<span id="page-26-2"></span>
$$
\rho_a = \pi \frac{\Delta \varphi}{I} \left( \frac{L(L^2 - a^2)}{a^2} \right). \tag{2-6}
$$

The Dipole-Dipole configuration is advantageous in the field for deep surveys. The electrodes of each pair a comparatively close to each other. Therefore, there is no need for long cables. This enables also surveys with very large offset and therefore greater depth penetration [\[Burger, 1992\]](#page-60-6).

Note that due to the principle of reciprocity, in all configurations the current and potential electrode pairs can be interchanged arbitrarily [\[Lowrie, 2007\]](#page-62-0).

#### <span id="page-26-0"></span>2-1-4 Inversion

The [ERT](#page-16-4) measurements give potential differences from which the apparent resistivities can be calculated easily by using Equation [2-3.](#page-24-0) For obtaining the true resistivities of a heterogeneous subsurface, an inversion of the data is required. The general concept of inversion is to generate synthetic data from a reference model and to compare it to the observed data. Based on the differences, the reference model is then adapted. This process is repeated iteratively until the reference model can explain the observed data within a predefined error tolerance. The generation of the synthetic data is done by forward modelling. The inversion problem is indeed unique but ill-posed [Günther et al., 2006]. This means in the case of [ERT,](#page-16-4) that there

can be multiple resistivity distributions, which result in the same observed data. A common inversion scheme is the Gauss-Newton scheme. In the following, the model parameters are denoted as  $\mathbf{m} = (m_1, m_2, ..., m_M)^T$  with M model cells and represent the resistivities in our model of the subsurface and the data vector  $\mathbf{d} = (d_1, d_2, ..., d_N)^T$  with N data points represents the observed apparent resistivities. The data contains errors from the measurement devices and inhomogeneities in the subsurface, which can be used for weighting, if they are available, which is usually not the case. With the aid of an  $L^2$ -norm of the weighted residual between the model response  $f(m)$  and the data, the data functional  $\Phi_d$  have to be minimized. The data functional  $\Phi_d$  is defined by [Günther et al., 2006]

<span id="page-27-0"></span>
$$
\Phi_d(\mathbf{m}) = \sum_{i=1}^N \left| \frac{d_i - f_i(\mathbf{m})}{\epsilon_i} \right|^2 = \left| \left| diag\left(\frac{1}{\epsilon_i}\right) (\mathbf{d} - \mathbf{f}(\mathbf{m})) \right| \right|^2_2.
$$
\n(2-7)

The errors  $\epsilon$  can either be measured or have to be estimated.

The minimization problem in Equation [2-7](#page-27-0) is highly ill-posed. Therefore, regularization should be applied by installing the model functional  $\Phi_m$ , which is defined as [Günther et al., 2006]

$$
\Phi_m(\mathbf{m}) = ||\mathbf{C}(\mathbf{m} - \mathbf{m}^0)||_2^2.
$$
\n(2-8)

Note that  $\mathbf{m}^0$  represents the reference model and  $\mathbf C$  the constraint or smoothness matrix. After to the insertion of  $\Phi_m$ , the weighted sum with the weighting factor  $\lambda$  needs to be minimized [Günther et al., 2006]

$$
\Phi = \Phi_d + \lambda \Phi_m \to \min. \tag{2-9}
$$

The weighting factor  $\lambda$  is also known as the regularization parameter. Its choice is a trade-off between data fit and model roughness. The higher the value of  $\lambda$ , the smoother is the model, but also the weaker is the data fit. [Günther et al., 2006]

At this point two options are possible. Either the reference model  $\mathbf{m}^0$  contains a priori information and  $C$  is a diagonal weighting matrix, then the model is not altered significantly from  $\mathbf{m}^0$ . Or  $\mathbf{m}^0$  is taken as a constant, while **C** handles model characteristics. The latter option means to apply smoothing. Because the problem is underdetermined, the latter option is the method of choice and therefore used in this project  $[G\ddot{u}]$  and  $[G\ddot{u}]$ .

In the case of regular meshes, the constraint matrix  $C$  can be taken as a discrete approximation of a partial differential operator of first or second order. If an irregular mesh is used, it is not that straightforward, because also neighbouring relations need to be considered. Therefore, the constraint matrix entries for the tetrahedra i and j with the common face f are  $C_{f,i} = -A$ and  $C_{f,j} = A$  with A as the face area. In total, the constraint matrix  $\mathbf{C} \in R^{B \times M}$  is sparse and contains  $2M$  non-zero values.  $B$  is the number of faces.

The model vector  **is updated in the manner of [Günther et al., 2006]** 

<span id="page-27-1"></span>
$$
\mathbf{m}^{k+1} = \mathbf{m}^k + \tau^k \Delta \mathbf{m}^k. \tag{2-10}
$$

 $\tau$  denotes the line search parameter, while k stands for the number of iteration. The line search parameter has to be recalculated in every iteration to avoid overshooting. In the Gauss-Newton scheme of minimizing  $\Phi$ , Equation [2-10](#page-27-1) transforms to [Günther et al., 2006]

<span id="page-28-1"></span>
$$
(\mathbf{J}^T \mathbf{D}^T \mathbf{D} \mathbf{J} + \lambda \mathbf{C}^T \mathbf{C}) \Delta \mathbf{m}^k = \mathbf{J}^T \mathbf{D}^T \mathbf{D} (\mathbf{d} - \mathbf{f}(\mathbf{m}^k)) - \lambda \mathbf{C}^T \mathbf{C} (\mathbf{m}^k - \mathbf{m}^0),
$$
 (2-11)

with **D** being the diagonal matrix  $diag(\frac{1}{\epsilon})$  $\frac{1}{\epsilon_i}$ ). Furthermore,  $\mathbf{J} \in R^{N \times M}$  denotes the Jacobian or sensitivity matrix. It is given by [Günther et al., 2006]

$$
\mathbf{J}_{i,j}(\mathbf{m}^k) = \frac{\partial f_i(\mathbf{m}^k)}{\partial m_j}.
$$
\n(2-12)

There are several methods for obtaining the sensitivity matrix **J** [\[McGillivray and Oldenburg, 1990\]](#page-62-10). But the [BERT](#page-16-3) software, which I apply for the inversion in this thesis, uses the formulation [\[Kemna, 2000\]](#page-61-9)

$$
\frac{\partial \varphi}{\partial \sigma} = -\frac{1}{\sigma I} \sum_{i} \sum_{j} A_{i,j} \varphi_i^{source} \varphi_j^{receiver}, \qquad (2-13)
$$

which is based on the reciprocity theorem.  $\varphi$  stands for the potential at the source and receiver positions. A denotes the sparse element matrix.  $\sigma$  is the electrical conductivity or the inverse of the resistivity. Special care needs to be taken to the elements adjacent to the electrodes, because there would an infinite potential occur. These need to be substituted by finite ones for restricting the sensitivity [Günther et al., 2006].

The described calculation needs to be done in every iteration. To reduce the computational effort, the solution of Equation [2-11](#page-28-1) can be solved approximately using conjugate gradient techniques [Günther et al., 2006].

The forward operator  $f(m)$  from Equations [2-7](#page-27-0) and [2-11](#page-28-1) needs to be recalculated in every iteration. This done by forward modelling. Several discretization methods can be implemented, but the most common are [FD](#page-16-5) and [FE.](#page-16-2) Here in this thesis, [FE](#page-16-2) will be applied and therefore explained in the next Section [2-2.](#page-28-0)

### <span id="page-28-0"></span>2-2 Finite elements method

When solving Partial Differential Equations [\(PDEs](#page-16-8)) analytically, several problems are encountered. First, it is very difficult to solve the problem for complex structures. So, it is helpful to divide the domain in smaller subdomains or cells, instead of solving the corresponding Partial Differential Equation [\(PDE\)](#page-16-8) once for the whole domain. This, furthermore, allows the investigation of local features. Additionally, a large matrix, which size depends on the number of cells, needs to be inverted. This is only possible, if the matrix is symmetrical. But often this is not the case. And even if it is invertible, this process requires an enormous computational effort. These issues can be circumvent by solving such problems numerically instead of analytically. But numerical methods are not capable of determining the exact solution for the [PDE,](#page-16-8) instead they give an approximation to the exact solution.

There exist different numerical techniques, e.g. the [FD](#page-16-5) method, the finite volume [\(FV\)](#page-16-9) method and the [FE](#page-16-2) method. Numerical methods are necessary for computer-based calculations of [PDEs](#page-16-8), because computers are not capable of calculating continuous problems.

<span id="page-29-0"></span>

Figure 2-3: Mesh comparison. a) Structured cubic mesh with local refinement. b) Unstructured tetrahedral mesh with local refinement. After [Rücker et al., 2006]

Hence, numerical discretizations are necessary. For this project, the [FE](#page-16-2) method is applied, because it is naturally capable of handling complex geometries [Rücker et al., 2006]. The [FE](#page-16-2) method can be applied in various ways. A common [FE](#page-16-2) technique is the Galerkin method, which I will describe exemplary in the following.

The first step in the [FE](#page-16-2) discretization is the meshing. The design of the mesh, which is used for the computation, is important. The number of nodes defines the spatial resolution and has therefore an significant impact on the accuracy. Furthermore, the number of nodes has a large contribution on the degrees of freedom, which determine the computational effort. For every node an equation needs to be solved. So, the more nodes are used, the more equations need to be solved. But an increased number of nodes also increases the accuracy. Hence, there is a trade-off between the accuracy and the computational effort in the mesh design.

The simplest mesh has structured quadratic/cubic elements, but they are unable to resolve complex geometries. With quadratic/cubic elements only right angles are realizable and therefore curvatures are very difficult to represent in the model. Furthermore, it can be helpful to refine the mesh at areas of interest to enhance the resolution in important areas, e.g. areas with strong gradients [Rücker et al., 2006]. When using quadratic/cubic meshes a local refinement is not efficient. It needs to be performed on whole rows and columns instead on some chosen elements [\[R¨ucker et al., 2006\]](#page-62-1). Therefore, the number of nodes increases significantly. Compared to quadratic/cubic elements, triangular/tetrahedral elements are more adaptable to complex geometries. This adaptability can be increased even more by the use of unstructured elements, because then arbitrarily angles can be used for representing curvatures. Additionally, the advantage of triangular/tetrahedral elements allows arbitrary local refinement. Due to the arbitrary angles, elements can have varying sizes. Summarizing, the use of unstructured triangular/tetrahedral meshes is advantageous, compared to quadratic/cubic meshes, and therefore they will be used in all meshes used in this project. For illustration, the comparison of a cubic and a tetrahedral mesh is given in Figure [2-3.](#page-29-0)

Additionally, the position of the grid points, the so-called nodes, need to be defined. The simplest option would be to set a node at each angle of the elements. But it is also possible to employ more nodes, e.g. at the center of the element edges or in the center of the element itself. Various combinations are imaginable. If multiple variables are computed, e.g. velocity and pressure for transport problems, it is possible to allocate certain nodes to a certain variable. But note, that the choice of number of nodes and their position has an impact on the stability of the outcome [\[Donea and Huerta, 2003\]](#page-61-10).

The [PDE](#page-16-8) that needs to be solved for the [ERT](#page-16-4) problem is [Rücker et al., 2006]

<span id="page-30-0"></span>
$$
\nabla \cdot \left(\frac{1}{\rho} \nabla u\right) = -\nabla \cdot \mathbf{j} \quad in \ \Omega \subset \mathbb{R}^3,
$$
 (2-14)

with the boundary condition [Rücker et al., 2006]

<span id="page-30-1"></span>
$$
\frac{1}{\rho} \left( \frac{\partial u}{\partial n} + \alpha u \right) = \mathbf{j} \cdot \mathbf{n} \quad on \Gamma. \tag{2-15}
$$

Note that  $\Omega$  is the domain, while  $\Gamma$  represents the boundary. u is the variable that is solved for, so in this case the electric potential and  $\mathbf{j}$  is the source current density. The **n** is the normal vector, which outwards directed on the corresponding boundary.

For solving the [PDE](#page-16-8) in Equations [2-14](#page-30-0) and [2-15](#page-30-1) with the [FE](#page-16-2) method, it needs to be transformed into the integral or weak formulation. This is done by multiplying the [PDE](#page-16-8) with the test function v and by integration over the domain  $\Omega$  [\[Donea and Huerta, 2003\]](#page-61-10). The transformed [PDE](#page-16-8) is given by [Rücker et al., 2006]

<span id="page-30-2"></span>
$$
\int_{\Omega} \frac{1}{\rho} \cdot \nabla v \nabla u \, d\Omega + \int_{\Gamma} \frac{1}{\rho} \alpha v u \, d\Gamma = \int_{\Omega} v I \delta(\mathbf{r} - \mathbf{r}_s) \, d\Omega + \int_{\Gamma} v \mathbf{j} \cdot \mathbf{n} \, d\Gamma. \tag{2-16}
$$

In Equation [2-16,](#page-30-2) v represents the test function, I denotes the source current and  $\delta$  is the Dirac delta function with r being the current position and  $\mathbf{r}_s$  the source position.

The integrals are hindering, because solving them analytically is very inefficient. Thus, they can be approximated with the aid of the Gauss quadrature integration. It approximates the integrals with a weighted sum (see Equation [2-17\)](#page-30-3) [Schwarz and Köckler, 2011]. The Gauss quadrature integration works more efficient, when applied to a reference element using the coordinates  $(\xi, \kappa, \zeta)$ . The reference element needs to have the same geometrical body as the physical elements, i.e. for a tetrahedral element, the reference element is also a tetrahedron [\[de Orio, 2001\]](#page-61-0). An illustration of a physical element and its corresponding reference element is depicted in Figure [2-4.](#page-31-1) Now, the Gauss quadrature integration transforms integrals of the [FE](#page-16-2) formulation in the sense of [\[Bellman and Casti, 1971\]](#page-60-7)

<span id="page-30-3"></span>
$$
\int_{-1}^{1} f(\xi) d\xi \approx \sum_{i=1}^{n_{quad}} w_i f(\xi_i). \qquad (2-17)
$$

Note that  $n_{quad}$  denotes the number of quadrature points and w the weighting factors. Equation [2-17](#page-30-3) is valid for the 1D case, but can be extended easily to the 2D or 3D case. The quadrature points are interpolated points within the reference element, which are required for the Gauss quadrature integration. The number, position and weighting  $w_i$  of the quadrature points depends on the type of element.

Furthermore, the basis or shape functions need to be defined. They interpolate the solution within the discrete nodes defined in the meshing process [\[de Orio, 2001\]](#page-61-0). Hence, with their

<span id="page-31-1"></span>

Figure 2-4: Tetrahedral finite element. Left the physical element in the original coordinate system. Right the reference element in the reference coordinate system. From [\[de Orio, 2001\]](#page-61-0)

aid a solution at every point within an element can be obtained. It is also advisable to define them on the reference element instead of the physical element. The basis functions  $N$  are typically chosen as low order polynomial functions, e.g. [\[de Orio, 2001\]](#page-61-0)

$$
N_i(\xi,\kappa,\zeta) = a_i + b_i\xi + c_i\kappa + d_i\zeta. \tag{2-18}
$$

The next step is to transform the reference coordinates to the physical coordinates  $(x, y, z)$ . When this is done, the approximate solution  $u_{\delta}$  can then be expressed with the aid of the basis functions [\[Donea and Huerta, 2003\]](#page-61-10)

<span id="page-31-2"></span>
$$
u_{\delta}(\mathbf{x}) = \sum_{A \in \nu \setminus \nu_D} N_A(\mathbf{x}) u_A + \sum_{A \in \nu_D} N_A(\mathbf{x}) u_D(\mathbf{x}_A), \tag{2-19}
$$

A denotes the node number, while the index  $D$  denotes the values corresponding to a Dirichlet boundary. The notation  $\nu\backslash\nu_D$  represents the all nodes excluding boundary nodes and  $\nu_D$  all nodes on the boundary [\[Donea and Huerta, 2003\]](#page-61-10). When combining Equation [2-19](#page-31-2) with the [PDE](#page-16-8) a system of equations is obtained for each element. The systems of equations of each element can be assembled to a global system of equations for the whole domain. This resulting system of equations is [\[Donea and Huerta, 2003\]](#page-61-10)

$$
Au = f, \t(2-20)
$$

where  $A$  is called the stiffness matrix and  $f$  the load vector. The solution is contained in the u vector. The system of equations then needs to be solved for u to obtain the solution.

### <span id="page-31-0"></span>2-3 Reduced basis method

For cases where a problem needs to be solved multiple times, solving the full problem every time can be very time consuming. Therefore, a method, which is able to lower the dimensionality of the problem, while keeping its accuracy on an appropriate level, is needed. A suitable technique is the [RB](#page-16-1) method. Such problems are, for example, uncertainty quantifications, where the uncertainties of input parameters are characterized, propagated forward through the model and assessed statistically. Another common application of the [RB](#page-16-1) method is the real time processing of experimental measured data.

In the following Section [2-3-1](#page-32-0) the concept of the [RB](#page-16-1) method is pointed out. The technique is split into two stages: The offline and the online stage. In the offline stage the reduced basis space is generated for the particular problem. It is an expensive computation, but has to be performed only once. A detailed description of the offline stage is given in Section [2-3-2.](#page-33-0) During the online stage, which is explained in Section [2-3-3](#page-35-0) the problem is solved multiple times using the previously generated reduced basis space. The calculation of a single solution in the online stage is then usually faster than using [FE.](#page-16-2) Due to the expensive offline stage, the [RB](#page-16-1) method is only reducing the computational effort, if many solutions are needed.

#### <span id="page-32-0"></span>2-3-1 Principles of the reduced basis method

The main interest is to find the exact solution  $u(\mu) \in \mathbb{V}$  such that [\[Hesthaven et al., 2016\]](#page-61-1)

$$
a(u(\mu), v; \mu) = f(v, \mu), \quad \forall v \in \mathbb{V}.
$$
\n
$$
(2-21)
$$

Note that  $\mu$  denotes the parameter that is varied, while v is the test function.  $a(u(\mu), v; \mu)$  and  $f(v, \mu)$  are the bilinear forms of the stiffness matrix A and the load vector f. But the exact solution is often mathematically impossible to obtain or requires an extreme computational effort. Thus, it is approximated by seeking the approximated solution [\[Hesthaven et al., 2016\]](#page-61-1)

$$
a(u_{\delta}(\mu), v_{\delta}; \mu) = f(v_{\delta}, \mu), \quad \forall v_{\delta} \in V_{\delta}, \tag{2-22}
$$

Properties with the index  $\delta$  are related to the approximation, in the project the [FE](#page-16-2) approx-imation. To simplify, it is assumed that the [FE](#page-16-2) approximated solution  $u_{\delta}(\mu)$  converges to the exact solution  $u(\mu)$  within a defined tolerance for any parameter  $\mu \in \mathbb{P}$  with  $\mathbb{P}$  being the parameter space. Therefore, the approximate solution can also be called the truth solution. Another assumption is that a low number of adequately selected basis functions is able to represent the truth solution well [\[Hesthaven et al., 2016\]](#page-61-1). Furthermore, it is assumed that the reduced basis space is a subset of the truth solution space. This also implies that  $N_{rb} \ll N_{\delta}$ , with  $N_{\delta}$  as the dimension of the truth space  $\mathbb{V}_{\delta}$  and  $N_{rb}$  as the dimension of the reduced basis space  $\mathbb{V}_{rb}$ .

The reduced basis problem is then stated as [\[Hesthaven et al., 2016\]](#page-61-1)

$$
a(u_{rb}(\mu), v_{rb}; \mu) = f(v_{rb}, \mu), \quad \forall v_{rb} \in \mathbb{V}_{rb}.
$$
\n
$$
(2-23)
$$

Note, that all properties denoted with the index rb correspond to the [RB](#page-16-1) method. For a proper implementation of the [RB](#page-16-1) method, the [PDE](#page-16-8) needs to be separated in a parameter-dependent and a parameter-independent part following the affine decomposition. Thus, the forms  $a(\cdot, \cdot; \mu)$ ,  $f(\cdot; \mu)$  and  $l(\cdot; \mu)$  need to be decomposed in the manner of [\[Hesthaven et al., 2016\]](#page-61-1)

$$
a(w, v; \mu) = \sum_{q=1}^{Q_a} \theta_a^q(\mu) a_q(w, v), \qquad (2-24)
$$

$$
f(v; \mu) = \sum_{q=1}^{Q_f} \theta_f^q(\mu) f_q(v), \qquad (2-25)
$$

$$
l(v; \mu) = \sum_{q=1}^{Q_l} \theta_l^q(\mu) l_q(v).
$$
 (2-26)

Note, that the l denotes the output functional. The forms  $a_q$ ,  $f_q$  and  $l_q$  are independent of the parameter  $\mu$  and the coefficients  $\theta_a$ ,  $\theta_f$  and  $\theta_l$  depend on  $\mu$ , but are independent of w and v. Because the forms  $a_q(\cdot, \cdot)$  do not depend on the parameter, a set of  $Q_a N_{rb} \times N_{rb}$ -dimensional matrices  $A_{rb}^q$  can already be precomputed after the generation of the reduced basis space in the offline stage. This allows usually a faster computation of  ${\bf A}_{rb}^{\mu}$  by [\[Hesthaven et al., 2016\]](#page-61-1)

$$
\mathbf{A}_{rb}^{\mu} = \sum_{q=1}^{Q_a} \theta_a^q(\mu) \mathbf{A}_{rb}^q.
$$
 (2-27)

The calculation of  $\mathbf{A}_{rb}^{\mu}$  in that manner is independent of  $N_{\delta}$ . The linear forms  $f(\cdot,\mu)$  and  $l(\cdot,\mu)$  are treated in a similar way.

If the affine decomposition is not applicable, the technique of empirical interpolation can help to determine an approximate form, which is suitable for the affine decomposition. For more information see [\[Hesthaven et al., 2016\]](#page-61-1).

Furthermore, it is important to note that the [RB](#page-16-1) method does not depend on a certain discretization method. Hence, it can be applied for example to the [FD,](#page-16-5) [FV](#page-16-9) or the [FE](#page-16-2) method.

Now, that the general concept of the [RB](#page-16-1) method is discussed, the question of accuracy arises. Since the [RB](#page-16-1) method approximates only the numerical approximation, the following triangle inequality holds [\[Hesthaven et al., 2016\]](#page-61-1)

<span id="page-33-1"></span>
$$
||u(\mu) - u_{rb}(\mu)||_{\mathbb{V}} \leq ||u(\mu) - u_{\delta}(\mu)||_{\mathbb{V}} + ||u_{\delta}(\mu) - u_{rb}(\mu)||_{\mathbb{V}}.
$$
 (2-28)

Above, it is assumed that the truth solution  $u<sub>\delta</sub>$  converges to the exact solution u with an appropriate accuracy. Therefore, the accuracy of the [RB](#page-16-1) method depends mainly on the second term on the right hand side of Equation [2-28.](#page-33-1) Note, that the [RB](#page-16-1) method gives not the same efficiency increase for every problem. While some [PDEs](#page-16-8) work well with the technique, others are unsuitable and have no efficiency increase or, even worse, a loss of efficiency. The method is most suitable for parameterized [PDEs](#page-16-8), which contain one or more parameters that can be separated in a parameter-dependent and parameter-independent part. For further investigation of the compatibility of a [PDE](#page-16-8) with the [RB](#page-16-1) method, there exists the Kolmogorov N-width. For more details on this, find more information on [\[Hesthaven et al., 2016\]](#page-61-1).

In the following, the focus turns to the two stages of the [RB](#page-16-1) method: At first the offline stage (see Section [2-3-2\)](#page-33-0), followed by the online stage (see Section [2-3-3\)](#page-35-0).

#### <span id="page-33-0"></span>2-3-2 Offline stage

In the offline stage the reduced basis space  $V_{rb}$  is generated, which is supposed to have a dimension as small as possible, but large enough to approximate the [FE](#page-16-2) solution within

<span id="page-34-0"></span>Algorithm: The greedy algorithm tol,  $\mu_1$  and  $n = 1$ . Input: A reduced basis space  $V_{\text{rb}}$ . Output:  $1.$ Compute  $u_{\delta}(\mu_n)$  solution to (3.1) for  $\mu_n$  and set  $\mathbb{V}_{\text{rb}} = \text{span}\{u_{\delta}(\mu_1), \ldots, u_{\delta}(\mu_n)\}\$  $\overline{2}$ . For each  $\mu \in \mathbb{P}_h$ **a.** Compute the reduced basis approximation  $u_{\text{rb}}(\mu) \in V_{\text{rb}}$  defined by (3.3) for  $\mu$ **b.** Evaluate the error estimator  $\eta(\mu)$  $\overline{3}$ . Choose  $\mu_{n+1} = \arg \max \eta(\mu)$  $\mu \in \mathbb{P}_h$  $\overline{4}$ If  $\eta(\mu_{n+1}) > \text{tol}$ , then set  $n := n+1$  and go to 1., otherwise terminate.



a certain tolerance. If it is chosen too small, the results have unacceptable uncertainties. If it is chosen too large, the computation becomes inefficient. The reduced basis space is generated by the Greedy algorithm or Greedy basis generation. Its scheme is depicted in Figure [2-5.](#page-34-0) It is an iterative process, where in each iteration one basis function is added, which increases the precision of the set of basis functions. In each iteration one truth solution has to be computed to generate the corresponding basis function. An essential component of the Greedy algorithm is the estimation of the error  $\eta(\mu)$  to determine how the precision of the solution has changed by replacing the [FE](#page-16-2) approximation space  $\mathbb{V}_{\delta}$  by the reduced basis space  $V_{rb}$ . The error is estimated, because determining the actual error is often linked with a high computational effort. Hence, an estimation of the error is usually advantageous [\[Hesthaven et al., 2016\]](#page-61-1). Since the [RB](#page-16-1) method is also used in real-time applications, where verifications are often neglected for saving computational resources, the estimation of the error needs to be fast and reliable. The results of the approximation, using error estimations, have often an appropriate accuracy, because the estimators use usually large training sets with a fine sampling of parameters. Hence, the error is based on the discrete set of parameters  $\mathbb{P}_h$ instead of the continuous parameter space  $\mathbb P$  [\[Hesthaven et al., 2016\]](#page-61-1).

There are various ways to estimate the error. In the utilized software package *DwarfElephant* (for more information see Section [3-3-3\)](#page-42-0) the relative output bound is used. Therefore, it is also used for the simulations in this project. The relative output bound  $\eta_{s,rel}$  is defined as [\[Hesthaven et al., 2016\]](#page-61-1)

$$
\eta_{s,rel}(\mu) = \frac{||\hat{r}_{\delta}(\mu)||^2_{\mathbb{V}}}{\alpha_{LB}(\mu) s_{rb}(\mu)}.
$$
\n(2-29)

Note that  $\alpha_{LB}$  represents the lower coercivity bound and  $\hat{r}_{\delta}$  denotes the Riesz representation of  $r(v_{\delta}; \mu) = f(v_{\delta}; \mu) - a(u_{rb}(\mu), v_{\delta}; \mu)$ . Furthermore,  $s_{rb}$  stands for the output [\[Hesthaven et al., 2016\]](#page-61-1). This error estimator is the rigorous upper bound of the output of interest. For a mathematical derivation of the error estimator and for alternative options beside the relative output bound, see [\[Hesthaven et al., 2016\]](#page-61-1). With the aid of the error bound, the resulting error from every iteration of the greedy algorithm can be evaluated and the number of basis functions can be chosen properly.

In the course of the iterative basis selection, for the case that there is at the  $n$ -th iteration step an *n*-dimensional reduced basis space  $V_{rb}$ , the estimated model order reduction error is maximized by the next basis function  $u_{rb}(\mu_{n+1})$ . Therefore, [\[Hesthaven et al., 2016\]](#page-61-1)

$$
\mu_{n+1} = \arg \max_{\mu \in \mathbb{P}} \eta(\mu). \tag{2-30}
$$

Then,  $u_{\delta}(\mu_{n+1})$  is calculated to extend the reduced basis space  $V_{rb}$  $span\{u_{\delta}(\mu_1),...,u_{\delta}(\mu_{n+1})\}\.$  Hence, the neglected basis function with the largest impact on the solution will be included in the reduced basis space  $V_{rb}$ . This procedure is continued in a loop until the maximal estimated error falls below a predefined error tolerance. Thus, the algorithm picks the next sample point, where the estimated error  $\eta(\mu)$ is maximal. Hence, the reduced basis tries to be optimal in the maximum norm over P [\[Hesthaven et al., 2016\]](#page-61-1). Note that for the case of insufficient accuracy of an  $N_{rb}$ -dimensional reduced basis space, it can be extended by  $n$  more modes, which has the same result as generating it initially with  $N_{rb} + n$  basis functions.

The computational effort of the offline stage is dominated by the necessary step of solving the truth problem  $N_{rb}$  times [\[Hesthaven et al., 2016\]](#page-61-1). But so far no actual solution is obtained. This is done in the online stage, which is outlined in the following section.

#### <span id="page-35-0"></span>2-3-3 Online stage

Now that the reduced basis space  $V_{rb}$  is defined, the procedure can move on to the online stage of the [RB](#page-16-1) method. In the case of an ideal setting, the computational effort would be independent of the dimension of the truth solution space  $N_{\delta}$ . It would only depend on the number of basis functions  $N_{rb}$   $(N_{rb} \ll N_{\delta})$  [\[Hesthaven et al., 2016\]](#page-61-1).

Actually, for each calculation with a different parameter the reduced basis solution matrix  ${\bf A}^{\mu}_{rb}$  would need to be recalculated. Because the parameter  $\mu$  can be contained in the bilinear form  $a(\cdot, \cdot; \mu)$ , the recalculation would require the construction of the truth matrix  $\mathbf{A}_{\delta}^{\mu}$  $\frac{\mu}{\delta}$ . This would have the issue that  $\mathbf{A}_{\delta}^{\mu}$  $\delta$  depends on  $N_{\delta}$ . But the  $N_{\delta}$ -dependent computation can be circumvent by assuming that the [PDE](#page-16-8) can be separated in a parameter-dependent and a parameter-independent part according to the affine decomposition described in Section [2-3-1.](#page-32-0) The computational effort of the online stage depends mainly on  $N_{rb}$  instead of  $N_{\delta}$ . Since  $N_{rb} \ll N_{\delta}$ , this is usually more efficient than computing the truth solution for every parameter value [\[Hesthaven et al., 2016\]](#page-61-1). From a more practical point of view, in the online stage one can compute various solutions at a reduced cost, since the basis space generation is already precomputed in the offline stage and this basis space just needs to be combined with the parameter dependent part of the [PDE](#page-16-8) [\[Hesthaven et al., 2016\]](#page-61-1).

If only one solution is required, the [RB](#page-16-1) method is much more inefficient than standard methods like the [FE](#page-16-2) method, because the basis space generation in the offline stage is very expensive. But when the basis space is generated, the computation of a single solution is much faster. Hence, the [RB](#page-16-1) method becomes more and more efficient, the more solutions need to be calculated [\[Hesthaven et al., 2016\]](#page-61-1).

### <span id="page-36-0"></span>2-4 Uncertainty Quantification

Uncertainties are a common issue in surveys and simulations. To obtain a better understanding of the reliability of the data, an uncertainty quantification is a helpful tool. It characterizes the input uncertainties and forward propagates them through the computational model. Furthermore, statistical or interval assessments can be performed on the resulting responses [\[Adams et al., 2014\]](#page-60-0).

There exist several ways of performing an uncertainty quantification. One way is the Bayesian calibration, where the uncertain parameters are characterized by probability density functions. These probability density functions describe the permissible parameter values. It then incorporates data to estimate with the likelihood function how well the model and its parameters describe the data [\[Adams et al., 2014\]](#page-60-0). The specific likelihood function used in software in this thesis is based on Gaussian probability density functions [\[Adams et al., 2014\]](#page-60-0).

A common method of the Bayesian calibration is the [MCMC](#page-16-6) method. It estimates the expectation of a function with respect to a distribution. It bases on a Markov chain, where a random walker explores the search space and successively visits solutions with stable frequencies originating from the distribution. After a burn-in phase, the Markov chain allows the walker to reach a stationary regime [\[Vrugt et al., 2009\]](#page-62-8).

For the process of the walk through the search space, the MCMC algorithm performs trial moves from the current position of the Markov chain  $x_{t-1}$  to a new state z. The candidate location z is sampled from a proposal distribution, which depends on the current location. The candidate location is then tested by the Metropolis acceptance probability, which either accepts or rejects the candidate location. If it is accepted, it becomes the new current location and the Markov chain moves on to z. Otherwise, the Markov chain stays at the current location  $x_{t-1}$  and a new candidate location is tested. After multiple iterations, the [MCMC](#page-16-6) algorithm will reach a stationary regime, if existing for the particular problem, and stop [\[Vrugt et al., 2009\]](#page-62-8).

The classical [MCMC](#page-16-6) algorithm has the issue, that it often requires a lot of iterations to reach the stationary regime. Therefore, an alternative scheme evolved to reduce the required computational effort: The [DREAM](#page-16-7) scheme [\[Vrugt et al., 2009\]](#page-62-8). It estimates a distribution of crossover probabilities that prefers large jumps over smaller ones in every chain. Hence, the stationary regime is reached faster [\[Vrugt et al., 2009\]](#page-62-8). Furthermore, outlier chains can occur that deteriorate the performance. In the [DREAM](#page-16-7) algorithm, these outlier chains are removed [\[Vrugt et al., 2009\]](#page-62-8). Additionally, the [DREAM](#page-16-7) scheme uses a decreasing step size. Thus, at the beginning of the algorithm large steps are performed to approach the stationary regime and later the steps are smaller to reach the stationary regime properly without overshooting it [\[Vrugt et al., 2009\]](#page-62-8).

For this research project, the [DREAM](#page-16-7) algorithm is used.

### Chapter 3

# Methods

<span id="page-38-0"></span>For giving more comprehensible insights in how the results given in Chapter [4](#page-44-0) are obtained, the general workflow (see Section [3-1\)](#page-38-1) and a short description of the utilized software (see Section [3-3\)](#page-41-0) is presented in this chapter.

### <span id="page-38-1"></span>3-1 General workflow

For getting an idea of how the results in the Chapter [4](#page-44-0) are obtained, the general workflow will be presented here. The workflow is illustrated in Figure [3-1.](#page-39-0)

The first step is to a generate synthetic data sets on the base of a simple predefined model. This model is designed in the software package [BERT](#page-16-3) ([Rücker et al., 2006],  $[G\$ unther et al., 2006), which is described in Section [3-3-1.](#page-41-1) For the calculation, a mesh is required. Therefore, [BERT](#page-16-3) utilizes the tool Tetgen [\[Si, 2003\]](#page-62-12). The next step is to perform a forward simulation on the generated mesh to obtain the synthetic data set. This process is also done using the software [BERT.](#page-16-3) To do so, the electrode and measurement configuration has to be defined. The forward simulation gives then for every single measurement the resulting potential difference and calculates the corresponding apparent resistivity. Furthermore, noise is added. Since the synthetic data are used as a surrogate for measured data, the noise is important to let the synthetic data appear more as measured data. The inversion is also performed by the software package [BERT.](#page-16-3) The idea of this back-and-forth process is to induce uncertainty. The theory behind the inversion is described in the theory part in Section [2-1-4.](#page-26-0)

On the base of the inversion result, a new model is built using Gmsh [\[Geuzaine and Remacle, 2009\]](#page-61-11). These rebuilt models are then used in the next step to perform forward simulations with the [FE](#page-16-2) and [RB](#page-16-1) method. Therefore, I use the software package *DwarfElephant*, which bases on the simulation environment MOOSE [\[Gaston et al., 2009\]](#page-61-12). Both are described in Section [3-3-3.](#page-42-0) The forward simulations of the [FE](#page-16-2) and [RB](#page-16-1) method are supposed to give the same result within a defined tolerance.

For comparing these data sets with the data set of the [BERT](#page-16-3) forward simulation, a Python

<span id="page-39-0"></span>

Figure 3-1: Schematic illustration of the workflow of the research project.

<span id="page-40-1"></span>

Figure 3-2: A two layer case used as the initial model. The upper layer is shown in red, while the lower halfspace is colored in blue. The background is indicated in grey.

script is used to perform multiple simulations by calling *DwarfElephant*. Each simulation is performed with a different source position. When all simulations are done, the potential differences of each measurement is calculated and assembled the same way as in the [BERT](#page-16-3) data set. The actual comparison of the data sets is supposed to be done in an uncertainty quantification. Unfortunately, this could not be done, due to a lack of time. Therefore, the software package *Dakota* [\[Adams et al., 2014\]](#page-60-0) is utilized, which calls the *Python* script to simulate the [ERT](#page-16-4) data set. The uncertainty quantification is supposed to be done twice. Once using the [FE](#page-16-2) forward simulation and once using the [RB](#page-16-1) forward simulation. There are two purposes for the uncertainty quantification: On the one hand, the uncertainties induced by the inversion process are supposed to be investigated and on the other hand the computational effort of the [FE](#page-16-2) and the [RB](#page-16-1) method are supposed to be compared in a large computation.

### <span id="page-40-0"></span>3-2 Model

The initial model, which is used for the inversion, is presented in the following. For the purpose of comparing the performance of [FE](#page-16-2) and [RB](#page-16-1) simulations and investigating the uncertainty induced by the inversion, no special requirements to the model are given. Hence, any model is suitable. Therefore, a very simple two layer case is used. To limit the computing times in the uncertainty quantification, the model is in 2D. The domain has an extend of 200  $m$ in x-direction and a depth of 50 m in z-direction. The upper layer has a thickness of 1 m, while the lower layer is the underlaying halfspace. Both layers themselves are completely homogeneous. The interface between the two layers is horizontal. The geometry of the model is depicted in Figure [3-2.](#page-40-1)

Furthermore, the resistivities of the layers need to be defined. A resistivity of 10  $\Omega m$  is allocated to the upper layer and a resistivity of 100  $\Omega m$  is assigned to the lower halfspace. According to Figure [2-1,](#page-23-1) these values are plausible for geoscientific purposes. The resistivity value of 10  $\Omega$ m of the upper layer could possibly represent for example an alluvium or a clay, while the resistivity value of 100  $\Omega$ *m* of the lower halfspace is suitable for example for a limestone or a sandstone.

### <span id="page-41-0"></span>3-3 Utilized software

In the progress of this research, several software packages are used. In order to make the process more transparent for the reader, the various software packages will be described briefly in the following. At first, the [ERT](#page-16-4) related software [BERT](#page-16-3) is introduced in Section [3-3-1,](#page-41-1) followed by the meshing software Gmsh in Section [3-3-2.](#page-41-2) The Section [3-3-3](#page-42-0) focuses on the simulation framework MOOSE with its [RB](#page-16-1) application DwarfElephant. Finally, the parametric analysis framework Dakota is introduced in Section [3-3-4.](#page-42-1)

#### <span id="page-41-1"></span>3-3-1 Boundless electrical resistivity tomography

The Boundless Electrical Resistivity Tomography software [Rücker et al., 2006][Günther et al., 2006], or abbreviated [BERT,](#page-16-3) focuses mostly on the inversion of direct current electrical measurements, but also allows forward modelling to obtain synthetic data. The used method is 'boundless' because the domain is extended to set the boundaries far from the area of interest to reduce the impact of boundary conditions. The software originated from the Direct Current Finite Element Method Library, or short DCFEMLib [Rücker et al., 2006], which is based on  $C_{++}$ . But over the time, it has been partly replaced by the Python-based *Generalized Inversion and Modelling Library* [Rücker et al., 2017], or short  $pyGIMLi$ .

The applied inversion scheme is generally based on the smoothness-constraint Gauss-Newton inversion. More details on the inversion scheme can be found in Section [2-1-4.](#page-26-0) The forward modelling, which is also part of the inversion process, is done using the [FE](#page-16-2) discretization method [Günther et al., 2006].

[BERT](#page-16-3) employs unstructured tetrahedral meshes. This has the advantage over structured cubic meshes that arbitrary geometries can be adapted sufficient and local mesh refinements are more efficient. For the mesh generation, [BERT](#page-16-3) employs the software  $TetGen$  [\[Si, 2003\]](#page-62-12). Compared to other software packages with a similar use, [BERT](#page-16-3) utilizes a triple grid approach instead of a dual grid approach. The parameter mesh is coarse and defines the resolution of the inversion. The secondary field mesh is globally refined and prolonged for the forward calculation. Finally, the primary potentials are computed on the very fine primary field mesh. The triple grid approach, including the secondary field mesh, allows a fast calculation [Günther et al., 2006].

#### <span id="page-41-2"></span>3-3-2 Gmsh

The software Gmsh [\[Geuzaine and Remacle, 2009\]](#page-61-11) is a 3D [FE](#page-16-2) grid generator. The intention of the Gmsh is to supply a fast, light and user-friendly meshing tool. It is designed to be fast enough to generate useful meshes for [FE](#page-16-2) calculations on standard personal computers and being light in the sense of required computational resources. Furthermore, the software is intended to be user-friendly by supplying a graphical user interface and having a robust and portable code. It contains a built-in CAD engine and postprocessor. Additionally, Gmsh features parametric input and advanced visualization capabilities. The software is completely developed in C++ [\[Geuzaine and Remacle, 2009\]](#page-61-11). Gmsh is an open-source software and can be obtained from <http://www.gmsh.info>.

#### <span id="page-42-0"></span>3-3-3 MOOSE Framework & DwarfElephant

The multiphysics object-oriented simulation environment, or short *MOOSE Framework* [\[Gaston et al., 2009\]](#page-61-12) was coupled with the Dakota Framework [\[Adams et al., 2014\]](#page-60-0). It is developed by the *Idaho National Laboratory* with the intention to provide a framework for scientist and engineers to develop new simulation applications with low effort. MOOSE is based on finite element [\(FE\)](#page-16-2) method using the continuous or discontinuous Galerkin formulation or the Petrov-Galerkin formulation. Furthermore, unstructured meshes with any dimensionality up to three is supported. The software is free to use and more information can be found on <https://mooseframework.org>.

The software package *DwarfElephant* is an application based on the *MOOSE Framework*. It offers simulations using the reduced basis [\(RB\)](#page-16-1) method. The application was developed by the CGRE Institute of RWTH Aachen University. It is free to use and available on Github on <https://github.com/cgre-aachen/DwarfElephant>.

#### <span id="page-42-1"></span>3-3-4 Dakota framework

The Dakota Framework [\[Adams et al., 2014\]](#page-60-0), developed by the Sandia National Laboratories in Albuquerque, NW, USA, is a software to perform advanced parametric analyses. It is no simulation software itself, instead it is aiding simulation software with giving additional options to analyse and improve the simulation. Despite containing internal simulation software, the Dakota Framework can be coupled with external simulation software. A scheme of the coupling is depicted in Figure [3-3.](#page-42-2)

Then a variety of iterative analyses can be performed. The software offers parameter studies and sensitivity analyses, which varies to input parameters to evaluate their impact on the simulation results. Another capability of the *Dakota Framework* are Design of Experiments techniques, which seek to extract trend data from a parameter space using a limited number of sample points. Furthermore, uncertainty quantifications are also part of the software. They forward propagate uncertainties in the input parameters trough the computational model and perform statistical and interval assessments on the responses. Additionally, the Dakota

<span id="page-42-2"></span>

Figure 3-3: Basic scheme of the coupling of the Dakota Framework with the user's simulation software. From [\[Adams et al., 2014\]](#page-60-0).

Framework offers optimization solvers, which are intended to maximize or minimize the objective function calculated by the coupled simulation software to optimize the performance and efficiency of the computation. Moreover, performing a calibration is also a possibility of the software. It is used to maximize the agreement between simulated data and measured or desired data. This is usually helpful for inverse problems. Because such iterative calculations are often consuming a lot of computational resources, the Dakota Framework is designed to exploit the capabilities of parallel computing [\[Adams et al., 2014\]](#page-60-0).

Summarizing, the *Dakota Framework* is created to obtain improved or even optimal experimental designs, to get a better understanding of the sensitivity and uncertainty of a specific problem and to generally to improve the performance of the simulation [\[Adams et al., 2014\]](#page-60-0). The Dakota Framework is free to use and more information can be found on [https:](https://dakota.sandia.gov) [//dakota.sandia.gov](https://dakota.sandia.gov).

### Chapter 4

### Results

<span id="page-44-0"></span>In the following, the results including intermediate results will be presented. In the first Section [4-1,](#page-44-1) I will describe the generation of the synthetic data and their corresponding inversion results. Afterwards I will give the model, which is generated on the base of the inversion result, followed by the forward simulation using the [FE](#page-16-2) and [RB](#page-16-1) methods in Section [4-2.](#page-48-0) The next section illustrates the results of the uncertainty quantification (see Section [4-3\)](#page-55-0).

### <span id="page-44-1"></span>4-1 Inversion results

After generating the model, which is described in Section [3-2,](#page-40-0) it is forward modelled to ob-tain synthetic field data with the aid of [BERT](#page-16-3) ([Rücker et al., 2006], [Günther et al., 2006]). Therefore, the position of the electrodes and their measurement configuration need to be defined. 18 electrodes are employed at the surface from x-positions ranging from 0 m to 17 m with an electrode spacing of  $1 \, \text{m}$ . The number is electrodes is rather small, but sufficient as the model contains only a feature at a depth of  $1 \, m$ , no deep penetration is required. For the measurements, the dipole-dipole configuration is chosen. The use of the Dipole-Dipole configuration allows more measurements for the same amount of source electrode positions than the Wenner and the Schlumberger configuration. For each source position, one forward simulation is required, since a new source position generates a different electrical field. Therefore, with the Dipole-Dipole configuration, more information can be obtained from a single forward simulation.

The distance between the two electrodes of the source and potential electrode pairs is kept constant of 1 m. Thus, the 18 electrodes allow 15 possible source positions without performing reciprocity measurements. For all 15 source positions, all possible measurements are performed. This results in a number of 120 measurements with offsets ranging from 2  $m$  and 16 m.

After the forward simulation an error of 3 % is added to the measured potential differences. This is import, because otherwise the synthetic data would give a perfect representation of the model. But synthetic data are not supposed to be perfect, because they should be surrogates for real data. Hence, the added error represents random noise in real measurements.

<span id="page-45-0"></span>

Figure 4-1: Pseudosection of the forward modelled apparent resistivities. Note that the z-axis only shows depth levels and no actual depth.

<span id="page-45-1"></span>

Figure 4-2: Pseudosection of the error in the forward simulation. Note that the z-axis only shows depth levels and no actual depth.

<span id="page-46-0"></span>

Figure 4-3: Resulting potential differences of the [BERT](#page-16-3) forward simulation. Note that each block represents all measurements with one source position, but varying offsets. Each block starts with a peak.

The resulting pseudosection of the forward modelling is depicted in Figure [4-1.](#page-45-0) The data are apparent resistivities. The actually modelled data is the potential, but with the potential differences of the potential electrodes the apparent resistivities can be calculated with the aid of Equation [2-6.](#page-26-2) The apparent resistivity values range from 10.14  $\Omega m$  to 62.11  $\Omega m$ . Furthermore, a gradient in the apparent resistivities can be observed. In the upper depth levels the smaller values are located, while the larger values are positioned in the lower depth levels. Additionally, the values within the depth levels vary. These variations are small in the upper depth levels and become more significant with depth. This is related to the error of the forward simulations shown in Figure [4-2.](#page-45-1) The error increases with the geometrical factor and therefore the depth level and so the measurements have more variations and are less reliable. The behaviour of the error depicts reality, because the current needs to penetrate more material and therefore more errors might be induced. The error distribution does not show more other features. For a later comparison of the forward simulation using [BERT](#page-16-3) and the forward simulation using *DwarfElephant*, I will use the measured potential differences. Alternatively, also the apparent resistivity could have been used. But none of these modelled properties gives more information as the other one. The resulting potential differences are depicted in Figure [4-3.](#page-46-0) Note that each block represents all measurements with varying offsets with one source position. Each block starts with the smallest possible offset and ends with the largest. The block become smaller with increasing measurement number, since the number of measurements per source position decreases to skip reciprocity measurements. The blocks follow a certain structure. Each block is similar to the previous one, but with the last measurement missing. Hence, measurements with same offsets give similar results. I will compare this potential difference data set later with the ones of the [FE](#page-16-2) and [RB](#page-16-1) data sets calculated with *DwarfElephant*. Please note that, in the following, I will distinguish the simulations of DwarfElephant by their method, namely [FE](#page-16-2) and [RB,](#page-16-1) and I will call the simulation done in

<span id="page-47-0"></span>

Figure 4-4: Resisitivity distribution as the result of the inversion for the two layer case. The electrode positions are indicated by the blue dots at the surface.

<span id="page-47-1"></span>

Figure 4-5: Interpreted resistivity distribution. One interface is interpreted an at depth varying from about  $0.8$  to about  $1.2$  m. Therefore, the interpretation suggests a two layer case.

[BERT](#page-16-3) only [BERT](#page-16-3) simulation, despite that it is a [FE](#page-16-2) simulation too.

In the next step, the synthetic data are inverted. This process is also performed with the [BERT](#page-16-3) software using the inversion scheme described in Section [2-1-4.](#page-26-0) This back-and-forth calculation is necessary because one of the ideas of this project is to investigate the uncertainty induced by the inversion. The inversion process ends after only two iterations, but the results are sufficient. The process is stopped, because the data fit criteria is reached by the error weighted  $\chi^2$  fit drops below 1. Its value is 0.9725. The relative root mean square between the data and the response is 10.264 %. Good values of the relative root mean square would be between  $1 - 2\%$  [Günther and Rücker, 2017]. Hence, the value here is not good but acceptable. The result of the inversion is given in Figure [4-4.](#page-47-0) In the inversion result, the upper area exhibits quite stable values. This area ranges from the surface to a depth varying from about 0.8 m to 1.2 m. The resistivity values in this upper area range from about 7.7  $\Omega$  to about 17.3  $\Omega$ m. After this depth, the resistivity increases stronger to a value of about 87.8  $\Omega$ m at a depth of about 5 m. Due to this change, an interface between two geological layers is interpreted. Hence, the interpreted interface varies from depths of about 0.8 m to 1.2 m. The interpretation is depicted in Figure [4-5.](#page-47-1) So, the result of the inversion suggests two layers. This correlates with the initial model, which also has two layers with the interface at  $1 \, m$ . But the interpreted interface is not straight, instead is has variations according to variations in the inversion result. The interpretation is used to build up a new model, which is used for the forward simulations using [FE](#page-16-2) and [RB](#page-16-1) in the software packages MOOSE [\[Gaston et al., 2009\]](#page-61-12) and DwarfElephant.

### <span id="page-48-0"></span>4-2 Forward modelling results

For the forward modelling with the [FE](#page-16-2) and [RB](#page-16-1) method using the software packages MOOSE [\[Gaston et al., 2009\]](#page-61-12) and DwarfElephant, a model and a mesh are required. I generated these on the base of the interpretation of the inversion result. The model with the corresponding mesh is displayed in Figure [4-6](#page-49-0) and a close-up on the area of interest is depicted in Figure [4-7.](#page-49-1) For the model, I used the 'boundless' concept of [BERT](#page-16-3) ([Rücker et al., 2006], [Günther et al., 2006]). Hence, the domain is expanded so the boundaries are placed far away from the region of interest with the intention to reduce the impact of the boundary conditions. Therefore, the model domain has a horizontal extent of  $200 \ m$  and a vertical extent of 50  $m$ . But note that the model is non-dimensionalized to an horizontal extent of 1 and a vertical extent of 0.25. This enhances the performance of the forward simulation, because the values for the calculation have similar magnitudes. The interface in the area of interest originates from the interpretation of the inversion results. But the inversion result gives no information about the depth of the interface at the boundaries. Therefore, I placed the interface at the boundaries to the average depth of the other interface points. The resistivity values of the two layers are obtained by averaging. The resulting resistivity of the upper layer is defined as 11.898  $\Omega m$  in the model, while a resistivity of 56.933  $\Omega m$  is assigned to the underlying halfspace.

For the mesh unstructured triangular elements are used. The element sizes vary intentionally to increase the resolution in the area of interest and decrease the resolution in less interesting regions. The mesh has 17266 elements and 8634 nodes in total.

Now, that the mesh is generated the forward simulations can be performed. First, I will present the results of the [FE](#page-16-2) simulations. Afterwards I will show the results of the [RB](#page-16-1) simulations.

The [ERT](#page-16-4) problem is a steady state problem, because a direct current is used and therefore, the electrical field is assumed to be constant over time. This can also be observed in the [ERT](#page-16-4) problem formulation in Equations [2-14](#page-30-0) and [2-15,](#page-30-1) which do not contain any time derivatives. The surface is considered as a Neumann boundary condition with a value zero. The air is very resistive and therefore, it is assumed that there is no flow through the surface. To the other boundaries Robin or Mixed boundary conditions are assigned. At these boundaries the electrical field can propagate through the boundaries. This represents that in reality the electrical field propagates into surrounding rocks. Therefore, the Robin boundaries have a stronger Neumann character and their penalty value is set to a small value of 1. Due to the 'boundless' concept, it is assumed that the boundary elements are far away and that their boundary conditions have no significant impact on the electrical field originated from the source. The source and sink at the source electrode pair are set to 0.1 A and  $-0.1$  A respectively. Note that the source and sinks are always positioned to a depth of 0.1  $m$  to avoid an interchange of the source and sink themselves with the Neumann boundary condition

<span id="page-49-0"></span>

Figure 4-6: Model and mesh based on the inversion result. Note that the domain is expanded following the boundless concept. Furthermore, the element sizes of the mesh vary to increase the resolution in the area of interest. The model and the mesh are generated with the aid of Gmsh. Note that the mesh is dimensionless to improve the performance of computations. The physical model would have a horizontal extent of  $200$  m and a vertical extent of  $50$  m.

<span id="page-49-1"></span>

Figure 4-7: Close-up of the model and its mesh based on the inversion result. The close-up shows the region of interest, which originates from the interpretation of the inversion, displayed in Figure [4-5.](#page-47-1)

<span id="page-50-0"></span>

Figure 4-8: Potential distribution of the [FE](#page-16-2) forward simulation exemplary for the first source position at 0 and 1 m. The upper layer has a resistivity of 11.898  $\Omega m$ , while the lower layer has a resistivity of  $56.933 \Omega m$ .

at the surface. For one full [ERT](#page-16-4) simulation along the whole profile, one forward simulation for each source electrode position needs to be performed. Therefore, 15 forward simulations are computed with the same electrode positions as in the [BERT](#page-16-3) simulation. Exemplary, the potential distribution of the first forward simulation with the source electrode positions at 0 and  $1 \, m$  is depicted in Figure [4-8.](#page-50-0) Furthermore, a close-up of the area of interest is shown in Figure [4-9.](#page-51-0) At first sight, the potential distribution behaves as expected. The potential has in maximum and minimum at the source and sink positions, respectively and decreases in magnitude with further distance from them. But despite that the source and sink inject and eject the same current of 0.1 A and  $-0.1$  A, the resulting magnitude of the positive maximum and negative minimum are not identical. While the maximum has a value of  $+0.956699 V$ , the minimum has a value of  $-0.943885$  V. Since they are both located in the same layer with the same resistivity, I do not expect a difference in magnitude.

Furthermore, the potential differences between the potential electrodes of each measurement are calculated and depicted in Figure [4-10.](#page-51-1) For purposes of comparing, all the three data sets of the [BERT,](#page-16-3) [FE](#page-16-2) and [RB](#page-16-1) simulations are compared in Figure [4-14.](#page-54-1) When comparing the data sets of [BERT](#page-16-3) and [FE,](#page-16-2) one notices that both data sets follow the same pattern. For each source electrode position, the potential difference decreases with offset. But the magnitude of the data sets vary. While the [BERT](#page-16-3) data set ranges from  $4.169 \times 10^{-4}$  V to  $5.662 \times 10^{-2}$  V, the [FE](#page-16-2) data set using *DwarfElephant* ranges from  $5.478 \times 10^{-3}$  V to  $8.676 \times 10^{-2}$  V. Furthermore, the gradient is steeper for the [FE](#page-16-2) data set than in the one simulated in [BERT.](#page-16-3)

Now, I will change my focus on the [RB](#page-16-1) simulations. They require more preparation. As described in Section [2-3-1,](#page-32-0) the [PDE](#page-16-8) needs to be split up in an parameter-dependent and parameter-independent part. The parameters I am varying are the resistivities  $\rho$  of the two layers. I performed the affine decomposition on the weak formulation of the [PDE](#page-16-8) given in Equation [2-16.](#page-30-2) Note that the volume integrals need to decomposed separately from the

<span id="page-51-0"></span>

Figure 4-9: A close-up on the area of interest of the potential distribution of [FE](#page-16-2) forward simulation exemplary for the first source position at  $0$  and  $1$   $m$ . The upper layer has a resistivity of  $11.898 \Omega m$ , while the lower layer has a resistivity of  $56.933 \Omega m$ .

<span id="page-51-1"></span>

Figure 4-10: Resulting potential differences of the FE forward simulation using *DwarfElephant*. Note that each block represents all measurements with one source position, but varying offsets. Each block starts with a peak.

surface integrals, since the volume integrals are part of the stiffness matrix and the surface integrals are part of the load vector. For both separations, one notices that the only parameter-dependent property is the parameter itself, hence the resistivity  $\rho$ .

After the affine decomposition, the offline stage can be performed. Therefore, a training set of parameters need to be defined. I defined the upper bound of the training set by adding 15 % to model resistivities, while I subtracted 15 % from the model resistivities for the lower bounds. Furthermore, an error tolerance between the truth solution and the reduced basis approximation needs to be defined. I set it to a value of  $1 \times 10^{-5}$ . The calculation of the offline stage gives a reduced basis space with 4 basis functions.

Please note that I encountered an error in my simulation in the last days of the project, which remains unsolved until the submission. The error prevents the [RB](#page-16-1) simulations from executing only the online stage with a precomputed reduced basis space. Therefore, for each forward simulation, the reduced basis space needs to computed again in the offline stage. Thus, the described forward simulation of the [ERT](#page-16-4) data set in the following is less efficient than the [FE](#page-16-2) forward simulation.

Since the [RB](#page-16-1) simulation is an approximation to the [FE](#page-16-2) simulation, I used the same simulation specifications. Hence, the same mesh is used, the same boundary conditions are applied with the same values and the source terms are identical. For comparing the results with the [FE](#page-16-2) simulations, the modelled potential distribution with the source and sink located at 0 and 1 m is x-direction is shown in Figure [4-11.](#page-53-0) Additionally, a close-up on the area of interest is depicted in Figure [4-12.](#page-53-1) The [RB](#page-16-1) simulation gives similar results as the one computed with the [FE](#page-16-2) method. The electrical field has its maximum and minimum at the source and sink, respectively, and the magnitude of the field decreases with the distance from the source and sink. But by comparing the actual values, one notices differences beyond the defined tolerance. The potential minimum has a value of  $-0.944155 V$ , while the maximum is 0.956866 V. For the minima, the error between the [FE](#page-16-2) and the [RB](#page-16-1) simulation is  $2.7 \times 10^{-4}$  V and  $1.67 \times 10^{-4}$  V. Since the error tolerance is defined to be  $1 \times 10^{-5}$ , there needs to be systematic error in the simulations.

Furthermore, the potential difference data set of the [RB](#page-16-1) forward simulation is depicted in Figure [4-13.](#page-54-0) The [RB](#page-16-1) data set is also compared in the plot in Figure [4-14](#page-54-1) with the data sets of the other simulations. The data set of the [RB](#page-16-1) simulation is similar to the one using the [FE](#page-16-2) simulation. But note that the potential differences in the [RB](#page-16-1) data set is for all measurements slightly smaller than the ones of the [FE](#page-16-2) data set.

Nevertheless, an important aspect of the [RB](#page-16-1) method and my project is the computational efficiency. Since I did not achieve a large computation for comparing the computing times of the [FE](#page-16-2) and [RB](#page-16-1) method, I compare the run times of single forward simulations. Using the [FE](#page-16-2) method, one forward simulation takes 1.120091 s. The calculation time of the [RB](#page-16-1) method needs to be distinguished in the run times of the offline and the online stages, because the offline stage usually needs to be calculated only once, while the online stage gives the run time of a single simulation when reduced basis space precomputed in the offline stage. The online stage takes  $1.0228 \times 10^{-2}$  s, while the offline stage requires 9.77001 s The gives a speed-up of 109.512, which is calculated by dividing the time of the [FE](#page-16-2) simulation by the time of the online stage. Furthermore, the pay-off, which gives the number of simulations, when the use of the [RB](#page-16-1) method is faster than the simulations of the [FE](#page-16-2) method. The pay-off has a value of 8.803. This calculations assume that the offline stage only needs to be calculated once.

<span id="page-53-0"></span>

Figure 4-11: Potential distribution of the [RB](#page-16-1) forward simulation exemplary for the first source position at 0 and 1 m. The upper layer has a resistivity of 11.898  $\Omega m$ , while the lower layer has a resistivity of  $56.933$   $\Omega m$ .

<span id="page-53-1"></span>

Figure 4-12: A close-up on the area of interest of the potential distribution of [RB](#page-16-1) forward simulation exemplary for the first source position at 0 and 1  $m$ . The upper layer has a resistivity of 11.898  $\Omega m$ , while the lower layer has a resistivity of 56.933  $\Omega m$ .

<span id="page-54-0"></span>

Figure 4-13: Resulting potential differences of the RB forward simulation using DwarfElephant. Note that each block represents all measurements with one source position, but varying offsets. Each block starts with a peak.

<span id="page-54-1"></span>

Figure 4-14: Comparison of the potential differences of the forward simulations using [FE](#page-16-2) (yellow) and [RB](#page-16-1) (green) with the aid of DwarfElephant and the forward simulation of [BERT\(](#page-16-3)blue).

### <span id="page-55-0"></span>4-3 Uncertainty quantification results

Unfortunately, no uncertainty quantification has been calculated. One of the original purposes of the uncertainty quantifications in this project was to investigate the uncertainties induced by the inversion process. Due to the error, which prevents the [RB](#page-16-1) simulation from executing the offline stage only, the uncertainty quantification using [RB](#page-16-1) would be meaningless. But an uncertainty quantification using [FE](#page-16-2) method could have been performed, but there was no time left. With more time, I could have at least find answers regarding the question of induced uncertainties.

# Chapter 5

### **Discussion**

<span id="page-56-0"></span>The major objective of the project is to investigate if and how the [RB](#page-16-1) method can be used for [ERT](#page-16-4) simulations to increase their computational efficiency. Therefore, I compared the simulations using the [RB](#page-16-1) method with simulations using the [FE](#page-16-2) method. These simulations are supposed to be used in an inversion and an uncertainty quantification for examining the required computational effort. Furthermore, the uncertainty quantification has the purpose of surveying the induced uncertainties by the inversion process in [BERT.](#page-16-3) Unfortunately, the time of the project ran out and I was unable to perform neither the inversion nor the uncertainty quantification. Furthermore, during the last days of the project, an error became apparent in the [RB](#page-16-1) problem. This error prevents the online stage from being executed only. This makes my [RB](#page-16-1) simulation less efficient than the [FE](#page-16-2) simulation. In the following, I will discuss the results and the problems occurred during the project.

The generation of the synthetic data set in [BERT](#page-16-3) and the following inversion performed both quite well, but since the initial model is rather simple, I did not expect major problems in the process. The data sets obtained from the [BERT](#page-16-3) and the [FE](#page-16-2) forward simulation show the same trend. With increasing offset between the source and the potential electrodes, the potential differences decrease. From a physical perspective, I expect such a behaviour. Since the potentials themselves decrease with the distance, the differences become smaller to. But the gradient of this trend differs for the two data sets. For smaller offset, the potential differences in the [BERT](#page-16-3) have a stronger gradient than in the [FE](#page-16-2) data set, but for larger offsets the gradient of the [FE](#page-16-2) data set becomes stronger than the gradient of the [BERT](#page-16-3) data set. These differences may occur, because the mathematics behind the simulations vary. For example, the [BERT](#page-16-3) simulation uses an approach including the calculation of the total potential and a secondary potential, while my [FE](#page-16-2) simulation only calculates the total potential. Furthermore, different models are used, which contribute to the different outcome. While the [BERT](#page-16-3) simulation uses the initial model, which I presented in Section [3-2,](#page-40-0) the [FE](#page-16-2) simulation uses the model, which is generated on the base of the inversion result. The initial model has resistivities of 10  $\Omega$ m and 100  $\Omega$ m for the upper and lower layers, respectively, while the resistivities obtained from the inversion process are 11.898  $\Omega m$  and 56.933  $\Omega m$ for the upper and the lower layer, respectively. While the resistivity of the upper layer in the new model is quite similar to the one of the initial model, the resistivity of the lower layer is remarkably smaller. Furthermore, there occur differences in the magnitudes of the potential differences. They may occur due to different source and sink currents. The currents are not clearly defined in the published papers introducing the software package [BERT](#page-16-3) [Rücker et al., 2006][Günther et al., 2006]. Since there is no current, which is optimal for any model, I would expect the currents chosen in [BERT](#page-16-3) to have a value of 1 with a decimal power. For example, values like  $\pm 0.01$  A,  $\pm 0.1$  A or  $\pm 1$  A and no arbitrary value. For investigating this, I tested different currents. The best tested value appears to be  $\pm 0.066$  A. Since there are other dissimilarities in the simulations like the model and the mathematical approaches, I do not expect to recover exactly the same values. Trying to fit the data neglecting the impact of other factors, I decided that a value of  $\pm 0.1$  A appears to be more reasonable. Note that the current of  $\pm 0.066$  A does not fit the dissimilarities in the gradients. Another possible origin of the deviations between the [BERT](#page-16-3) and [FE](#page-16-2) data sets is an incorrectly implementation of the non-dimensionalized model lengths into the [PDE.](#page-16-8) If a source and sink current of  $\pm 1$  V is selected and the resulting [FE](#page-16-2) data are scaled down by the non-dimensionalization factor of 200, the data are in a similar range as the ones obtained from [BERT.](#page-16-3) This may be coincidental, but can also be the reason for the mismatch. Further error evaluation needs to be done here. An additional uncertainty quantification investigating the source and sink currents may give more insights.

Beside the dissimilarities between the data sets of [BERT](#page-16-3) and [FE,](#page-16-2) also the [FE](#page-16-2) and [RB](#page-16-1) simulations do not show the desired similarity. Despite an error tolerance of  $1 \times 10^{-5}$  is defined in the generation of the reduced basis space, the error between the [FE](#page-16-2) and [RB](#page-16-1) reaches values above  $1 \times 10^{-4}$ . I would expect the error to occur, because I have a systematic error in my problem formulation in the simulation software DwarfElephant. Inserting the decomposed [PDE](#page-16-8) correctly into the simulation is no straightforward task and I expect that I did not accomplish it appropriately.

At the end of the project, I discovered a conceptual inefficiency in my project. I only considered the resistivity as a varying parameter in the [RB](#page-16-1) method. But [ERT](#page-16-4) measurements require multiple source positions. Hence, it would be very helpful to implement the source position as a varying parameter too. With only the resistivity as a varying parameter, one reduced basis space would need to be generated for each source position. For my measurement configuration, the offline stage need to be calculated 15 times. This reduces the efficiency of the [RB](#page-16-1) method. Therefore, for a proper use of the [RB](#page-16-1) method, the source position needs to be a varying parameter too. So, in my approach the reduced basis space needs to be calculated 15 times, once for each source position. But that is only a workaround, I expect to be significantly less efficient than the one considering the source position as a varying parameter.

The uncertainty quantification has not been performed, due to the lack of time. It was supposed to investigate the resistivity values. Since the resistivities of the lower layer differ significantly in the two models, I also expect the obtained value from the inversion to be quite uncertain. However, the resistivity values of the upper layer are quite similar, so I expect the uncertainty quantification to evaluate it to be quite certain. Since the model of the two simulations differ, I expect also some minor uncertainties originating from these differences.

The speed-up of using the [RB](#page-16-1) method instead of the [FE](#page-16-2) method is with a value of 109.512 a remarkable improvement. Since the reduced basis space contains 4 basis functions, I expected a significant value. Also the pay-off of 8.803 is quite low. This pay-off would even reduce the computing time of a full [ERT](#page-16-4) data set using the rather small measurement configuration used in this project. For larger computations the use of the [RB](#page-16-1) method can reduce the computing time significantly.

# Chapter 6

# Conclusion

<span id="page-58-0"></span>Summarizing, I see a place of the reduced basis method (RB) in the practical application of the electrical resistivity tomography. The full simulation of RB measurements requires the calculation of multiple forward simulations, one for each source position. For my approach without varying the source positions, the offline stage needs to be calculated once for each source position, which limits the efficiency increase by the use of RB. Now, I would suggest to include the source position as a varying parameter in the RB method. Then, the offline stage would only need to be computed once and the computational efficiency would increase even more. This is especially advantageous, if the Wenner configuration is used, instead of the Dipole-Dipole configuration in this project. The Wenner configuration allows only a single measurement per source electrode position, while the Schlumberger and the Dipole-Dipole configuration allow multiple measurements. For large calculations like uncertainty quantifications or the inversion of field data, the expected reduction in computing time would be significant. But note that the use of the RB method requires more effort in the implementation. Hence, it always needs to be evaluated if the planned computation is large enough that the use of RB pays off. Unfortunately, I did not achieved my aims for the project to prove the full potential of the RB method for the ERT problem, by performing either an inversion or an uncertainty quantification using RB simulations. But hopefully my results will help in upcoming research to be more successful.

The next step of upcoming research would be to vary not just the resistivities, but also the source position in the RB. This would increase the potential of the technique even more. Additionally, the implementation of the RB method into the inversion process is an important step, since the inversion is a necessary processing step for data of most ERT field measurements. Furthermore, the RB method could be implemented in optimizing the field measurements by enabling an inversion of measured data directly in the field. Then the measurement configuration and parameters could be modified to improve the outcoming results.

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