

Four-Channel Integrated-Optic Wavelength Demultiplexer With Weak Polarization Dependence

Arjen R. Vellekoop and Meint K. Smit

Abstract—A new planar four-channel wavelength demultiplexer with weak polarization dependence based on an optical-phased array, is proposed and demonstrated. An experimental device with dimensions $4.5 \times 3.2 \text{ mm}^2$ is designed and fabricated, using conventional (high quality) optical lithography. The demultiplexer operates in the wavelength range 776.5–781.2 nm, with a channel spacing of 1.55 nm. Insertion loss was 0.6 dB for the central channels and 1.2 dB for the outer channels for TE polarization, excluding 1.3-dB waveguide propagation loss. Crosstalk values measured 15.4–29.7 dB for the TE and 13.4–22.2 dB for the TM polarization.

I. INTRODUCTION

OPTICAL wavelength division multi demultiplexing (WDM) is reckoned to be a useful technique to increase the transmission capacity of a fiber. The key components in a WDM transmission system are the wavelength multiplexers and demultiplexers, for which loss and crosstalk figures, channel spacing, device size, and polarization dependence are important properties.

Various configurations of planar (de)multiplexers have been accomplished by using optical interference filters [1]–[3], wavelength selective coupling [4]–[6], and optical diffraction gratings [7]–[9]. Verbeek *et al.* [1] reported a weakly polarization-dependent four-channel Mach-Zehnder demultiplexer (device length 33 mm) with a channel spacing of 7.7 nm, 2.6-dB loss, and 16-dB channel crosstalk. Rottman *et al.* [4] realized a polarization-independent dual-channel (de)multiplexer with 30-nm channel spacing, 3-dB loss, and 26–40-dB crosstalk, based on two-mode interference (TMI) and with a device length of 12 mm. Imoto *et al.* [5] proposed and demonstrated a dual-channel directional coupler (de)multiplexer with 100-nm channel spacing, 5-dB loss, and 20-dB crosstalk. The configuration was designed for a specific polarization and had a length of 10 mm. Suhara *et al.* [7] demonstrated a five-channel demultiplexer for multimode systems with 30-nm channel spacing using a chirped grating. The hybrid component, with dimensions $20 \times 13 \text{ mm}^2$, showed 9.5-dB loss and 21.5-dB crosstalk. Non-planar grating components using GRIN lenses [10], [11] achieve better performance. Such devices, however, need to be assembled with great precision and are therefore less attractive.

In this paper a new four-channel wavelength demultiplexer based on an optical phased array is reported. The demultiplexer is weakly polarization dependent and has insertion loss and

crosstalk figures comparable to the best results reported so far, but with a considerably smaller device size.

II. BASIC PRINCIPLE

Recently a planar optical phased array, consisting of a number of concentrically bent waveguides, with both focusing and dispersive properties was reported [12]. The component had considerable loss, however, evidenced by the occurrence of multiple foci in the focal plane, which is characteristic for phased arrays. The power coupled to higher order beams can be reduced by spacing the individual elements more closely. This can be achieved by providing the array of fan-in and fan-out coupling sections at both ends, as shown in Fig. 1. These coupling sections gradually adapt the in- and outgoing beams to the set of guided modes in the array. To obtain a smooth connection between the concentric section and the coupling section, an adapter section is required.

The phase transfer of the complete phased array (including coupling and adapter sections), can be controlled by choosing the radii R_i of the concentric section such, that the total length of each channel equals an integer number of wavelengths. This choice of the phase transfer will transform the divergent incoming beam into a convergent outgoing one with the same angular field distribution, so that the source field at the transmitter side will be reproduced in the focal plane at the receiver side. If the array is designed such that the length of the array channels increases linearly with their rank number, then a small variation of the propagation constant β will result in a linear variation of the phase transfer which will tilt the outgoing beam and thus lead to a lateral shift of the focal position.

Since the propagation constant β in a planar waveguide depends on the polarization as well as the wavelength, the phased array can in principle operate as a polarization splitter as well as a wavelength (de)multiplexer.

The periodic properties of the array function of the phased array have been used to realize a four-channel wavelength demultiplexer with weak polarization dependence.

III. THEORETICAL PROPERTIES OF THE DEMULTIPLEXER

The most important physical properties of the demultiplexer are the dispersive lateral focal displacement $d(\Delta\beta)$, the aberration $a(\Delta\beta)$ (shown in the enlargement of the focal plane in Fig. 1), and the insertion loss L_{\max} of the outer channels, which depend on the geometrical parameters of the array as derived in [13]

$$d(\Delta\beta) = f \cdot \Theta(\Delta\beta) = \frac{\Delta\beta}{\beta} \cdot \Psi \cdot \left(f + t + \frac{S_0}{2} \right) \quad (1)$$

Manuscript received September 11, 1989; revised October 2, 1990.

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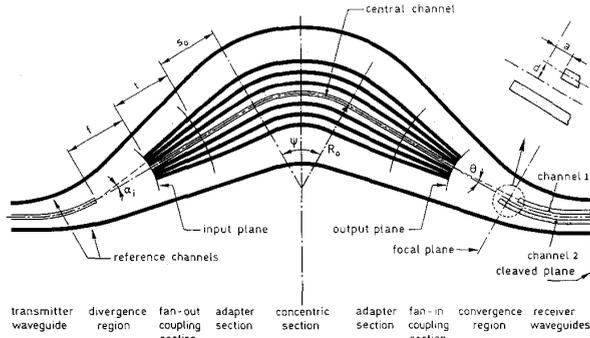


Fig. 1. Demultiplexer geometry. Only two reference channels and two receiver channels are drawn.

$$a(\Delta\beta) = 2 \cdot \frac{\Delta\beta}{\beta} \cdot \left(f + t + \frac{s_0}{3} \right) \quad (2)$$

$$\Theta_{\max} = \Theta_0 \cdot \sqrt{\left(\frac{L_{\max}}{20 \cdot \log_{10}(e)} \right)} \quad (3)$$

in which $\Delta\beta$ is the deviation of the propagation constant β from the central design value β_0 , Ψ is the concentric-section sector angle, f is the focal length, t is the fan-in and fan-out section length, s_0 is the adapter central-channel length, Θ_0 is the beam divergence half angle (at the $1/e^2$ point) of the single-waveguide far field, described by $\Theta_0 = \lambda/\pi \cdot w_0$ with λ the wavelength in the film and w_0 the $1/e^2$ modal half width, and Θ_{\max} is the maximum dispersion angle (the angular deviation relative to the optical axis as shown in Fig. 1) for which $L \leq L_{\max}$.

Since the propagation constant β is dependent on the polarization state of the incoming signal, the phase transfer of the phased array is polarization dependent. If the demultiplexer is designed such that the focal point coincides with the optical axis for the TE polarization at the central wavelength λ_0 , then, according to (1), the outgoing field for the TM polarization is tilted by an angle

$$\Delta\Theta_p = \Theta_{TE} - \Theta_{TM} = \frac{\Delta\beta_p}{\beta f} \cdot \Psi \cdot \left(f + t + \frac{s_0}{2} \right) \quad (4)$$

in which Θ_{TM} and Θ_{TE} represent the phase-front angles of the output field for the two polarizations and $\Delta\beta_p = \beta_{TE} - \beta_{TM}$.

Due to the periodic character of the array function, the phase transfer will be periodic in β . If the dispersion angle $\Delta\Theta_p$ equals the period of the array function, the phase transfer of the demultiplexer will be essentially polarization independent. The period Θ of the array function being described by $\sin^{-1}(\lambda/w) \approx \lambda/w$, with λ the wavelength in the film and w the wavelength width, polarization independence occurs if

$$\Delta\Theta_p = m \cdot \frac{\lambda}{w} = \frac{2\pi}{\beta w} \cdot m$$

$$m = \dots -2, -1, 0, 1, 2, \dots \quad (5)$$

in which m is an arbitrary integer number.

IV. DEMULTIPLEXER DESIGN

A single-mode AlGaAs laser, which is thermally tunable between 775–785 nm, was applied to measure the device, thus

fixing the design central wavelength at $\lambda_0 = 780$ nm. Designs at longer wavelengths may be realized equally well.

The same (bimodal) waveguide structure (lateral waveguide width $w = 2 \mu\text{m}$, optical contrast $\Delta n \approx 0.02$), which was successfully applied in a polarization splitter [13], was chosen for the demultiplexer. Due to the slightly longer wavelength (780 nm instead of 633 nm) the minimal bending radius is increased to 1 mm. The central propagation constant β_0 ($12.46 \mu\text{m}^{-1}$), the $1/e^2$ modal half-width w_0 ($1.34 \mu\text{m}$) and the difference in propagation constant $\Delta\beta_p$ ($0.17 \mu\text{m}^{-1}$) of the two polarizations β_{TE} and β_{TM} are fixed with the choice of the waveguide geometry.

As described in Section III, the demultiplexer can be made polarization independent by choosing the parameters such that the angle $\Delta\Theta_p$ and the array-function period coincide. It then follows from (4) and (5) that

$$f = \frac{d}{\left(\frac{2\pi}{\beta w} \right)} \cdot \frac{\Delta\beta_p}{\Delta\beta} \quad (6)$$

With a lateral receiver-waveguide spacing $d = 5 \mu\text{m}$ and a relative-channel spacing $\Delta\beta/\beta_0 = 0.22\%$, the focal length becomes $123 \mu\text{m}$.

Channel crosstalk considerations are analogous to the polarization splitter design described in [13]. A receiver-waveguide spacing of $5 \mu\text{m}$ and a number $N_p = 31$ array waveguide should be sufficient for a channel isolation better than -45 dB. The design parameters and corresponding theoretical properties are summarized in Table I.

It is shown in [13] that if the fan-in and fan-out section length t is chosen at least equal to the focal length, the mutual coupling between adjacent waveguides will become negligible. However, with a value $t = f = 123 \mu\text{m}$, the concentric section sector angle Ψ , as determined from (1), will be greater than 180° , which can lead to waveguide crossings. For Ψ to become 180° , t has to be chosen $460 \mu\text{m}$.

The first priority in the design considerations was to demonstrate that the phased array could perform a wavelength-demultiplexing function. The demultiplexer was therefore designed for TE polarization. The polarization-independence was achieved by adjusting the focal length according to (6). Consequently, the aberration correction of the receiver channels, as is pointed out in Fig. 1 and in (2), only applies to the TE polarization. The aberration for the TM field is considerable. By substituting $\Delta\beta = \Delta\beta_p = 0.17 \mu\text{m}^{-1}$ in (2), the aberration a_{TM} becomes $18 \mu\text{m}$. In comparison with a focal depth of $f_d = \pi \cdot w_0/\lambda = 5.4 \mu\text{m}$, being the distance for which the beam waist is less than $w_0 \cdot \sqrt{2}$ (Gaussian approximation), it can be expected that the demultiplexer properties for TM polarization will be worse than for TE polarization. To average the demultiplexer performances for both polarizations, the aberration correction should be averaged to $\frac{1}{2} \cdot (a_{TE} + a_{TM})$, so that the receiver waveguides are placed such that the focal mismatch is equal for both polarizations. This correction is not implemented for the present design. An even better solution would be to obtain an aberration-free design. Research on the feasibility of such a design is currently being performed.

The theoretical performance of this design can be computed by following the transmitted signal through the demultiplexer as described in [13]. The field at the input plane of the array is calculated as the two-dimensional diffraction field of the fundamental-mode profile at the end of the transmitter waveguide.

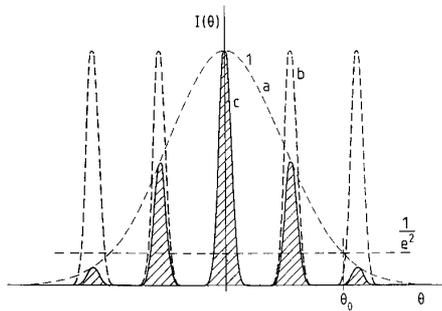


Fig. 2. Field-intensity distribution I as a function of the dispersion angle θ . The field intensity (pattern c) is formed by the product of the angular radiation pattern of the singular waveguides (pattern a), and the array function of the phased array (pattern b).

TABLE I
DESIGN PARAMETERS AND CALCULATED PROPERTIES OF THE
FOUR-CHANNEL WAVELENGTH DEMULTIPLEXER

| | |
|--|---|
| wavelength | $\lambda_0 = 780 \text{ nm}$ |
| central propagation constant (TE) | $\beta_0 = 12.46 \mu\text{m}^{-1}$ |
| propagation constant channel spacing | $\Delta\beta = 0.0274 \mu\text{m}^{-1}$ |
| difference ($\beta_{TE} - \beta_{TM}$) | $\Delta\beta_p = 0.17 \mu\text{m}^{-1}$ |
| $1/e^2$ modal half width | $w_0 = 1.34 \mu\text{m}$ |
| $1/e^2$ beam divergence half angle | $\theta_0 = 6.86^\circ$ |
| focal length | $f = 123 \mu\text{m}$ |
| central channel radius | $R_0 = 1200 \mu\text{m}$ |
| fan-in fan-out section length | $t = 460 \mu\text{m}$ |
| transceiver maximum angle | $\alpha_{\text{max}} = 14.0^\circ$ |
| number of waveguides | $N_p = 31$ |
| adapter central length | $s_0 = 300 \mu\text{m}$ |
| concentric-section sector angle | $\Psi = 177.8^\circ$ |
| dispersion angle channels 1, 4 | $\theta_{1,4} = \pm 3.5^\circ$ |
| dispersion angle channels 2, 3 | $\theta_{2,3} = \pm 1.16^\circ$ |
| displacement channel 1, 4 | $d_{1,4} = \pm 7.5 \mu\text{m}$ |
| displacement channel 2, 3 | $d_{2,3} = \pm 2.5 \mu\text{m}$ |
| aberration channel 1, 4 | $a_{1,4} = \pm 4.5 \mu\text{m}$ |
| aberration channel 2, 3 | $a_{2,3} = \pm 1.5 \mu\text{m}$ |
| insertion loss channel 1, 4 | $L_{1,4} = 2.26 \text{ dB}$ |
| insertion loss channel 2, 3 | $L_{2,3} = 0.25 \text{ dB}$ |
| channel crosstalk | $< -45 \text{ dB}$ |

Next, coupling of this field into the individual waveguides is computed. The propagation through the array is described by $\exp(-j\beta l_i)$, where l_i is the total channel length. At the output plane the field is reconstructed as the sum field of the individual channels from which the two-dimensional diffracted field in the focal plane is calculated (the error made in neglecting coupling effects is cancelled because the same error occurs at the input side). Aberration and truncation effects are inherently accounted for in this simulation.

Fig. 2 depicts the focal field-intensity distribution I as a function of the dispersion angle θ (pattern c), formed by the product of the angular radiation pattern of the singular waveguides (pattern a), and the array function (pattern b). As can be seen from the picture the field intensity decreases with increasing distance of the receiver to the optical axis. This insertion loss is described by (3).

The power transfer is determined by overlapping the field in the focal plane with the modal fields of the receiver waveguides. Results are shown in Fig. 3 as a function of wavelength. The simulation predicts an insertion loss below 2.5 dB and a crosstalk of -40 dB , which is 5 dB higher than the design value

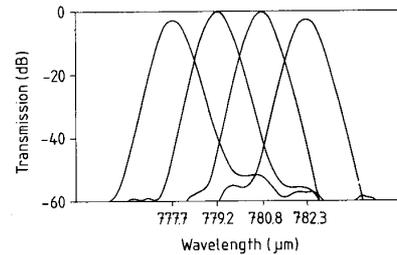


Fig. 3. The numerically simulated power transfer as a function of the wavelength λ .

due to inclusion of aberration and truncation effects (as a result of the finite aperture width).

As the demultiplexer is designed for a specific wavelength, deviations from the design wavelength (due to production tolerances or thermal effects) will decrease the demultiplexer performance. Variations within $\pm 10\%$ of the relative channel spacing $\Delta\beta/\beta_0$ have a negligible effect. With the above values, the acceptable relative propagation constant fluctuation lies within $\pm 10\%$ of $1.7 \text{ nm}/780 \text{ nm}$, i.e., $\pm 2 \cdot 10^{-4}$, corresponding to a maximum wavelength fluctuation $\Delta\lambda \leq \pm 0.17 \text{ nm}$.

V. FABRICATION

The experimental demultiplexer was fabricated in an RF sputter deposited $0.25\text{-}\mu\text{m}$ Al_2O_3 -layer ($n \approx 1.69$) waveguide structure on a thermally oxidized silicon substrate, as described in [14], [15]. The lateral waveguide structure in this layer is produced by atom-beam milling a 40-nm step and covering the structure with a sputtered $0.6\text{-}\mu\text{m}$ SiO_2 layer ($n \approx 1.46$), so that an embedded ridge guide structure is formed.

The $2\text{-}\mu\text{m}$ waveguide pattern is obtained by projecting an optical pattern generator (ASET COMBO 250) created chromium mask onto an image reversal resist film (Hoechst 5214 E) with a $4\times$ Canon reduction camera (FPA 141).

A photograph of the chromium mask is shown in Fig. 4. The two bent waveguides at the upper side and the two straight waveguides at the lower side of the phased array, are used as reference channels. By comparison of the simultaneously excited reference channels with the receiver channels, loss and crosstalk values can be determined. The array contains 31 waveguides. Device size is $4.5 \times 3.2 \text{ mm}^2$.

VI. RESULTS

The experimental device was investigated by coupling light into the device from an AlGaAs semiconductor laser operating around 780 nm with a prism coupler as described in [16]. Input and reference channels were excited simultaneously with a broad beam. Transmission loss was determined by comparing the output intensity to that of the reference channels. Maximum measurement errors, as determined from comparison of identical (straight) waveguides, are within $\pm 0.5 \text{ dB}$. A thermoelectric Peltier module was used to tune the semiconductor laser wavelength between 775 and 785 nm with a temperature coefficient $d\lambda/dT \approx 0.25 \text{ nm/K}$. The cleaved endface of the device, indicated in Fig. 1, is projected onto a CCD video camera with a microscope objective. The camera signal is digitized and processed by a computer. Camera observations at different wavelengths are shown in Fig. 5. Fig. 6 shows a horizontal intensity scan over the picture of Fig. 5(b), from which the insertion loss

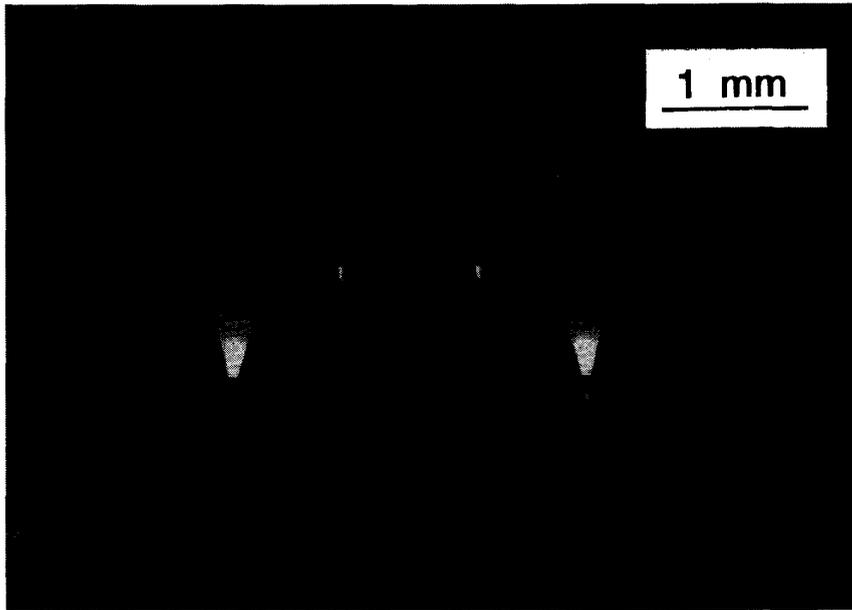


Fig. 4. Mask pattern of the demultiplexer. The upper two bent waveguides and the lower two straight waveguides are reference channels.

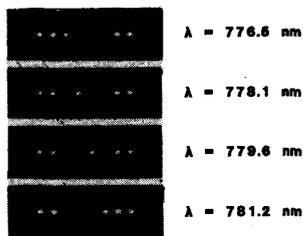


Fig. 5. Camera observations of the channels at the cleaved end face of the device for the four consecutive channel wavelengths.

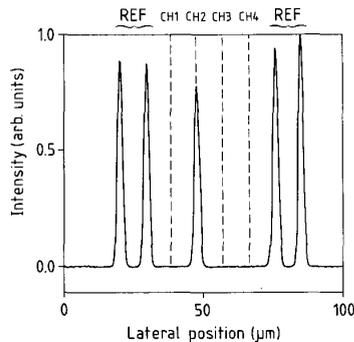


Fig. 6. Intensity scan for $\lambda = 778.1$ nm (TE polarization). Estimated insertion loss is 0.6 dB.

can be derived by comparing the intensity from the receiver channel with the bent reference channels (the left two reference channels). Referencing to the bent reference waveguides instead of the straight ones eliminates the effect of waveguide losses which can be seen to amount to 1.3 dB. This loss is fully contributable to the additional 8550- μm length of the curved

waveguides, and is in good agreement with the 1.5-dB/cm propagation loss measured in straight waveguides. For the TE polarization the so found insertion loss (excluding waveguide propagation loss) was 0.6 dB for the central channels and 1.2 dB for the outer channels, for the TM polarization the insertion loss measured 2.1 dB for the central and 3.2 dB for the outer channels.

The TE insertion loss for the outer channel is 1 dB lower than the theoretically expected value (3), Table IIa). This may be explained as follows. The limited resolution of the optical lithography will cause filling in of the interwaveguide gap for small gap widths, as can be seen from the photograph (Fig. 4). Due to this filling in, the actual length of the fan-in section will become shorter than the design value. The actual focal length will increase with the same amount.

In the expression for the lateral focal displacement $d(\Delta\beta)$ (1), the dispersion angle $\Theta(\Delta\beta)$ will reduce with increasing focal length f . The right-hand side of the expression will remain unchanged, however, because the total length $f + t$ will be unchanged. With a reduced dispersion angle $\Theta(\Delta\beta)$ the insertion loss L will reduce according to (3).

The length over which the gaps are filled is experimentally determined to be approximately 70 μm . With a corrected focal length $f = 193 \mu\text{m}$ the actual dispersion angle for the outer channels becomes 2.2°. It then follows from (3) that the maximum insertion loss $L_{\text{max}} = 0.9$ dB. The measured insertion loss of 1.2 dB for the outer channels appears to be in good agreement with this corrected value.

Channel isolation is determined by comparing the intensity of the receiver channels, after removing a 20-dB optical attenuator, to that of the reference channels, and measured 15.4–29.7 dB for the TE polarization and 13.4–22.2 dB for the TM polarization.

Table II lists the values measured for the demultiplexer. The laser wavelength was measured by coupling part of the input beam to a monochromator with ± 0.1 -nm resolution.

TABLE II
MEASUREMENT RESULTS OF THE DEMULTIPLEXER FOR TE- AND
TM- POLARIZATION
(Insertion loss and crosstalk values are determined relative to the
bent reference waveguides, eliminating waveguide losses.)

| Channel | Wavelength (nm) | TE transmitted power (dB) | | | |
|---------|--------------------|---------------------------|-------|-------|-------|
| | | 1 | 2 | 3 | 4 |
| 1 | 776.5 | -1.2 | -19.0 | -21.3 | -25.7 |
| 2 | 778.1 | -19.4 | -0.6 | -20.0 | -20.6 |
| 3 | 779.6 | -17.5 | -16.2 | -0.8 | -15.4 |
| 4 | 781.2 | -29.7 | -21.7 | -16.0 | -1.2 |

| Channel | Wavelength (nm) | TM transmitted power (dB) | | | |
|---------|--------------------|---------------------------|-------|-------|-------|
| | | 1 | 2 | 3 | 4 |
| 1 | 776.5 | -2.8 | -17.0 | -21.5 | -22.2 |
| 2 | 778.1 | -17.6 | -2.1 | -16.5 | -20.0 |
| 3 | 779.6 | -20.4 | -17.4 | -3.2 | -15.9 |
| 4 | 781.2 | -18.8 | -19.1 | -13.4 | -3.0 |

The insertion loss and channel isolation for the TM polarization are below the values for TE polarization, due to the large focal aberration occurring for this polarization.

VII. CONCLUSIONS

A novel four-channel wavelength demultiplexer with weak polarization dependence was proposed and demonstrated. The experimental device, realized using conventional high-quality optical lithography, measured an insertion loss (excluding 1.3-dB waveguide propagation loss) of 0.6 dB for the central channels and 1.2 dB for the outer channels for TE polarization, and 2.1 and 3.2 dB, respectively, for TM polarization. Channel isolation was 15.4–29.7 dB for TE and 13.4–22.2 dB for TM polarization, relative to the reference channels. Channel spacing was 1.55 nm. These properties are comparable to the best results reported so far, but realized with a considerably smaller device size.

ACKNOWLEDGMENT

The authors wish to thank J. W. M. van Uffelen and A. H. de Vreede for fabricating the devices, and Prof. G. A. Acket and Prof. B. H. Verbeek for comments and helpful suggestions.

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